$\rm INFN/AExxx, \, LAL-xxx, \, SLAC-R-xxx$

SuperB Detector Technical Design Report

Abstract

This report describes the technical design detector for Super B.

Contents

1	Intr	oduction	1
	1.1	The Physics Motivation	1
	1.2	The Super B Project Elements	1
	1.3	The Detector Technical Design Report	2
2	Acc	elerator Overview	7
3	Det	ector Overview	9
	3.1	Physics Performance	9
	3.2	Challenges on Detector Design	12
	3.3	Open Issues	14
	3.4	Detector R&D	14
4	Phy	sics with SuperB	19
	4.1	Introduction	19
	4.2	B and D decays	19
		4.2.1 Rare B decays	19
		4.2.2 Rare D decays	21
		4.2.3 CKM matrix and unitarity triangle	22
		4.2.4 CP violation in B decays $\ldots \ldots \ldots$	23
		4.2.5 CP violation in D decays $\ldots \ldots \ldots$	23
		4.2.6 Other symmetry tests	24
		4.2.7 Charm mixing	24
		4.2.8 <i>B</i> physics at the $\Upsilon(5S)$	25
	4.3	au physics at Super B	26
		4.3.1 Lepton flavor violation in τ decay	26
		4.3.2 <i>CP</i> violation in τ decay	27
		4.3.3 Measurement of the τ $g-2$ and EDM form factors	27
	4.4	SuperB Neutral Current Electroweak Physics Programme	28
	4.5	Exotic Spectroscopy in Super B	29
	4.6	Direct searches	31
	4.7	Executive Summary	31
5	Mad	chine Detector Interface and Backgrounds	37
	5.1	OverviewM.Sullivan, M. Boscolo E.Paoloni, - 1 page	37
	5.2	Backgrounds sources. M.Sullivan, M.Boscolo, E.Paoloni, - 2 pages	37
	5.3	Radiative Bhabha	37
		5.3.1 Simulation tools	37
		5.3.2 Losses at the beam-pipe	38
		5.3.3 Shield System	39
	5.4	Pairs ProductionC.Rimbault - 2 pages	40

	5.5	Touse	check bacgkround. M.Boscolo - 2 pages				•	40
	5.6	Beam	n gas background. M.Boscolo - 2 pages					40
	5.7	Synch	nrotron radiation background. M.Sullivan - 2 pages					40
	5.8	SVT b	background overview					40
	5.9	DCH	background overview R.Cenci D.Lindemann - 2 pages					43
	5.10	FTO	F background overview					43
	5.11	FDIR	C background overview					44
		5.11.1	Shielding the FDIRC					44
		5.11.2	Background rates in the FDIRC					45
		5.11.3	Integrated charges and doses					45
	5.12	EMC	background overview. S.Germani - 2 pages					45
	5.13	IFR b	background overview V.Santoro - 2 pages					46
	5.14	ETD	background overview					46
	5.15	SVT	radiation monitor. A.Di Ciaccio- 3 pages					47
	5.16	Quick	demounting. M.Sullivan, F.Bosi, E.Paoloni - 4 pages					47
		-						
6	Silic	on Ver	tex Tracker					51
	6.1	Overv	iew G.Rizzo - 12 pages	 •	• •		•	51
		6.1.1	SVT and Layer0				•	51
		6.1.2	SVT Requirements			•	•	53
			6.1.2.1 Resolution			•	•	53
			6.1.2.2 Acceptance				•	53
			6.1.2.3 Efficiency				•	55
			6.1.2.4 Background & Radiation Tolerance					55
			6.1.2.5 Reliability					56
		6.1.3	Baseline Detector Concept					56
			6.1.3.1 Technology			•	•	56
			6.1.3.2 Layout			•	•	56
			6.1.3.3 Electronic Readout				•	58
			6.1.3.4 Module design and Mechanical Support				•	60
		6.1.4	Layer0 Pixel Upgrade				•	61
			6.1.4.1 Motivations			•	•	61
			6.1.4.2 Technology Options for Layer0 pixel upgrade			•	•	62
			6.1.4.3 Pixel Module & Material Budget			•	•	64
		6.1.5	R&D Main Activities			•	•	65
	6.2	Backg	rounds R.Cenci - 4 pages	 •		•	•	65
		6.2.1	Pair production			•	•	66
		6.2.2	Radiative Bhabha				•	66
		6.2.3	Touschek			•	•	66
		6.2.4	Beam Gas			•		66
		6.2.5	Other sources	 •		•	•	66
	6.3	Detect	tor Performance Studies N.Neri - 6 pages					66
		6.3.1	Introduction					66
		6.3.2	The SVT layout					66
		6.3.3	Impact of Layer0 on detector performance					67
		6.3.4	Tracking performance	 •			•	70
		6.3.5	Impact of machine background on tracking performance					70

	6.3.6	Sensitivity studi	ies for time-depende	ent analyses			72
	6.3.7	Performance wit	th Layer0 pixel dete	ctors			73
	6.3.8	Particle identifie	cation with dE/dx				74
6.4	Silicor	Sensors	Ĺ	. Bosisio - 8 pages			74
	6.4.1	Requirements .					75
	6.4.2	Sensor design an	nd technology				75
	6.4.3	Wafer lavout an	d quantities				78
	6.4.4	Prototyping and	l tests				78
	6.4.5	z-side strip conr	ection options				78
6.5	Fanou	t Circuits	L.Vita	ale - M.Prest $2+2$ pages			79
0.0	651	Fanouts for lave	er()	20 1000 1000 Pages			80
	0.011	6511 Requir	rements				80
		6512 Techno	ology				80
		6513 Design					80
		6514 Protot	vning and tests				80
	652	Fanouts for out	or lavers				80
	0.0.2	6521 Requir	ements				80
		6522 Materi	al and production t	echnique			80
		6523 Dosign	ai and production i	echnique			80
		6.5.2.4 Testa s	and prototuping				80
66	Flootr	0.0.2.4 Itsis a	ind prototyping				80
0.0	6.6.1	Boadout chips		$V B_0 = 10$			82
	0.0.1	6611 Floatro	onia Pondout for Stu	vine - 10	•••••		04 82
	669	Poodout chips r	onic Readout for Str	ip and surplet Detecto			04 92
	6.6.2	Readout Chip I	mplementation				00 85
	6.6.4	Readout Onip in Readout Onip in	adout ching				86
	0.0.4	Hubrid Design	eadout cmps	M Cittorio 10			00 87
	0.0.0	Dete Transmissi	ion	$M_{\text{Cittorio}} = 10$			01
	6.6.7	Data Hansinssi Dawar Supply	1011	$\mathbf{M}.\mathbf{O}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}\mathbf{I}I$			01
67	0.0. <i>i</i>	Power Supply	d Aggembler	- 2	14 magaz		01
0.7	6 7 1	IP Constraint	d Assembly	5. Dettariii / F. Dosi -	14 pages		01
	0.7.1	I.R. Constraint					01
	0.7.2	Detector Assemb	ly				90
	0.7.3	C 7 2 1 I I I F	I alf detector accord				92
		0.7.3.1 L1-L3 6729 I0 Hal	fair detector assem	101y			92
		0.7.3.2 L0 пал 6.7.2.2 Монет	I detector assembly	TITE to W comical abi			90
		0.7.3.3 Mount	LU to be-pipe and	LI-LO to w conical sin	elding		90
		0.7.3.4 Installa	Demounting	Assembly into the Super	B Detector .		99
	674	Detector Discor	Demounting			1	99
	0.75	Detector Mac	ient and Survey			l 1	.02
	0.7.5	Detector Monito	oring			l	.02
		0.1.5.1 Positio	m Monitoring System	Ш MDI		l	.02
	0.7.0	0.7.5.2 Kadiat	non Monitoring: in .	WIDI		l	.03
	0.7.0	K&D Program				l	.03
		0.7.0.1 Cables	· · · · · · · · · · · · · · · · · · ·			l	.03
		0.7.0.2 Hybrid	1			l	.03
		6.7.6.3 Be Bea	am pipe			1	.03
		6.7.6.4 LU Mo	dule			1	.03

		6.7.6.5	Inner layer sextant	
		6.7.6.6	Arch modules	
		6.7.6.7	Cones and space frame	
		6.7.6.8	Full-scale model of IR and W (Cylinder Shield
		6.7.6.9	Quick Demounting test	
	6.8	Laver0 Upgrad	le Options G.Riz	zo/L.Ratti - 10 pages
		6.8.1 Techno	logy options	
		6.8.1.1	Hybrid pixels	
		6.8.1.2	Deep N-well CMOS monolithic	sensors 106
		6.8.1.3	Monolithic pixels in CMOS que	adruple well technology
		6.8.2 Overvie	ew of the B&D activity	107
		6821	Front-end electronics for hybrid	nixels in planar and 3D CMOS tech-
		0.0.2.1	nology	107
		6822	The Appel DNW MAPS series	109
		6.8.2.3	The Apsel/well quadruple well	monolithic sensor 112
		683 Badiati	ion tolerance	11 <i>A</i>
	6.0	Services and I	Itilities	- 1 pages 116
	0.3	6.0.1 Sorvice	and Utilities	- + pages
		6.0.1.1	Deta and control lines	116
		0.9.1.1 6 0 1 2	Data and control lines	
		6.0.1.2	Cooling water	116
		6.0.1.4	Dry air or nitrogen	117
		0.9.1.4	Dry an or introgen	
7	Drif	t Chamber		123
	71	Overview	- Finocchiaro	Roney 10 pages 123
		7.1.1 Physics	s Requirements	- 3 pages
		7.1.2 Geome	trical Constraints	123
		713 Machin	e Background Considerations	- Cenci 3 pages 123
		714 DCH D	esign Overview	- 2 pages 123
		715 Expect	ed Performance	- 2 pages 123
	7.2	Design Optimi	zation - Finocchiaro He	earty Piccolo Roney 9 pages 124
		7 2 1 Cluster	Counting	124
		7.2.2 Cell De	esign and Laver Arrangement	124
		7 2 3 Gas Mi	xture	126
		724 B&D a	nd Prototype Studies	126
		7241	Prototype 1	128
		7.2.1.1 7.2.4.2	Prototype 2	120
		7.2.4.2	Single Cell Prototypes	130
		7.2.4.0	A ging studies	130
		7.2.1.1 7.95 Dl-DE	utura Davelopments	135
				1.1.1
	7 2	Mechanical De	wign	125
	7.3	Mechanical De	sign	
	7.3	7.2.5 R&D F Mechanical De 7.3.1 Endpla	esign	
	7.3	7.2.5R&D FMechanical De7.3.1Endpla7.3.2Inner c7.3.3Outer f	sign	$ \dots $
	7.3	7.2.5R&D FMechanical De7.3.1Endpla7.3.2Inner c7.3.3Outer 07.3.4Chaire	sign	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$
	7.3	7.2.5R&D FMechanical De7.3.1Endpla7.3.2Inner c7.3.3Outer C7.3.4Choice7.3.5Endpla	sign	$ \begin{array}{cccccccccccccccccccccccccccccccccccc$

		7.3.6	Endplate systems
			7.3.6.1 Electronics enclosures
			7.3.6.2 Cooling
			7.3.6.3 Shielding
			7.3.6.4 Electromechanical boards
		7.3.7	Stringing
	7.4	Electr	ronics - Felici, Martin 1 page
		7.4.1	Design Goals
		7.4.2	Standard Readout - charge measurements specifications
			7.4.2.1 Resolution
			7.4.2.2 Dynamic range
			7.4.2.3 Linearity
		7.4.3	Standard Readout - time measurements specifications
			7.4.3.1 Resolution
			7.4.3.2 Dynamic Range
			7.4.3.3 Linearity
		7.4.4	Standard Readout - DCH Front End Electronics (overall design)
		7.4.5	Standard Readout - On Detector Electronics
			7.4.5.1 Preamplifier Boards
		7.4.6	Sampled Waveforms - specifications
			7.4.6.1 Resolution
			7.4.6.2 Dynamic range
			7.4.6.3 Linearity
		7.4.7	Sampled Waveforms - DCH Front End Electronics (overall design) 141
		7.4.8	Sampled Waveforms - On Detector Electronics
			7.4.8.1 Preamplifier Boards
	7.5	High	Voltage system- Robertson, Martin 1 page142
		7.5.1	Main HV system and cable routing- Robertson 0.5 page 142
		7.5.2	Distribution boards - Martin 0.5 page
		7.5.3	HV distribution boards - Standard ReadOut
	7.6	Gas s	ystem
	7.7	Calibr	ration and monitoring - Roney 3 pages
			7.7.0.1 Slow control systems
			7.7.0.2 Calibration
			7.7.0.3 Gas monitoring system
			7.7.0.4 On-line monitor
	7.8	Integr	ation
		7.8.1	Overall geometry and mechanical support
		7.8.2	Installation and alignment
		7.8.3	Services
	7.9	Biblio	graphy $\ldots \ldots 146$
8	Part	ticle Ide	entification 141
-	8.1	Sum	nary of Physics Requirements and Detector Performance goals
		8.1.1	Physics requirements
		8.1.2	Detector concept
		8.1.3	Charged Particle Identification

	8.2	Parti	le Identification Overview	2
		8.2.1	Experience of BABAR DIRC	3
		8.2.2	Barrel PID: Focusing DIRC (FDIRC)	3
	8.3	Proje	ted Performance of FDIRC 2-3 pages	7
		8.3.1	Reconstruction Arnaud, Roberts	7
		8.3.2	MC Simulation	7
		8.3.3	Effect of Background on performance Roberts	7
	8.4	The I	Barrel FDIRC Detector Overview	7
		8.4.1	Photodetectors	7
		8.4.2	Laser calibration system	7
		8.4.3	FDIRC Mechanical Design	9
		8.4.4	Electronics readout, High and Low voltage	9
		8.4.5	Integration issues	4
		8.4.6	FDIRC R&D Results until now	5
		8.4.7	Ongoing FDIRC R&D	8
		8.4.8	System Responsibilities and Management	3
		8.4.9	Cost, Schedule and Funding Profile	3
	8.5	A pos	sible PID detector on the Super <i>B</i> forward side	С
		8.5.1	Physics motivation and detector requirements	С
		8.5.2	Forward PID R&D activities	1
		8.5.3	The Forward task force	4
		8.5.4	The DIRC-like forward time-of-flight detector (FTOF)	3
9	Elec	tromag	netic Calorimeter 197	7
	9.1	Overv	ew	7
		9.1.1	Background and radiation issues	3
		9.1.2	Simulation tools	9
			9.1.2.1 Fastsim	9
			9.1.2.2 FullSim	9
	9.2	Barrel	Calorimeter)
		9.2.1	Requirements Relevant to the Super B Environment)
			9.2.1.1 Crystal Aging at $BABAR$)
			9.2.1.2 Backgrounds)
		9.2.2	Description of BABAR Barrel Calorimeter	1
			9.2.2.1 Mechanical design $\ldots \ldots 201$	1
			9.2.2.2 Readout	3
			9.2.2.3 Low-energy Source Calibration	5
			9.2.2.4 Light Pulser	7
		9.2.3	Performance of BABAR barrel	7
			9.2.3.1 Energy and position resolution	7
			9.2.3.2 Gamma-gamma mass resolution	7
			9.2.3.3 Radiation Damage Effects on Resolution	9
			9.2.3.4 Expected Changes in Performance at Super B	9
		9.2.4	Electronics changes)
			9.2.4.1 Requirements)
			9.2.4.2 Electronic noise measurements)
			$0.949 \text{D}_{\text{res}} = 1 \text{ and } 1 an$	a

	9.2.5	SLAC De-installation, Transport and Local Storage
	9.2.6	Electronics refurbishment
	9.2.7	Re-installation at Tor Vergata
9.3	Forwa	rd Calorimeter $\ldots \ldots \ldots$
	9.3.1	LYSO Crystals
		9.3.1.1 Introduction
		9.3.1.2 Optical and Scintillation Properties
		9.3.1.3 Radiation Hardness
		9.3.1.4 Specifications, Production and Testing
	9.3.2	Readout and Electronics
		9.3.2.1 APD Readout
		9.3.2.2 Electronics Design
		9.3.2.3 Requirements on mechanics
	9.3.3	Calibrations
		9.3.3.1 Initial LYSO calibration with source
		9.3.3.2 Electronics calibration
		9.3.3.3 Temperature monitoring and correction
	9.3.4	Mechanical Structure
		9.3.4.1 Crystals
		9.3.4.2 Modules
		9.3.4.3 Installation
		9.3.4.4 Refurbishment of the BaBar structure
		9.3.4.5 Spare FWD modules survey and tests
	9.3.5	Tests on Beam
		9.3.5.1 Description of apparatus
		9.3.5.2 Description of the beams
		9.3.5.3 Description of data and calibration
		9.3.5.4 Electronics noise measurements
		9.3.5.5 Temperature corrections
		9.3.5.6 Results
	9.3.6	Alternatives
		9.3.6.1 Full LYSO calorimeter
		9.3.6.2 Pure CsI
		9.3.6.3 BGO
		9.3.6.4 Comparison among options
9.4	Backw	ard Calorimeter
	9.4.1	Requirements
		9.4.1.1 Energy and angular resolution
		9.4.1.2 Background rates
		9.4.1.3 Radiation hardness
		9.4.1.4 Solid angle, transition to barrel
	9.4.2	Mechanical design
		9.4.2.1 Calorimeter construction
		9.4.2.2 Support and services
	9.4.3	SiPM/MPPC readout
	9.4.4	Electronics
	9.4.5	Calibration

		9.4.6	Backward simulation	246
		9.4.7	Performance in simulations	246
		9.4.8	Impact on physics results	246
		9.4.9	Use for particle identification	248
		9.4.10	Discussion of task force conclusions	251
	9.5	Trigger	e	252
		9.5.1	Calorimeter readout trigger	252
			9.5.1.1 Normal mode	252
			9.5.1.2 Calibration mode	252
		9.5.2	Calorimeter trigger primitives	252
	9.6	Enviro	nmental monitoring	252
10	Instr	rumente	ed Flux Return	259
	10.1	Physic	s Requirements and Performance Goals	259
	10.2	Detect	or Overview	259
		10.2.1	The Absorber Structure	259
		10.2.2	The Active Detector Choice	261
	10.3	Backg	rounds	261
		10.3.1	Main background sources	261
			10.3.1.1 Neutron Background	261
			10.3.1.2 Charged Particles	262
			10.3.1.3 Photon background	263
		10.3.2	Background remediation	263
		10.3.3	Radiation doses on the IFR detector	264
	10.4	Identif	ication Performances	264
		10.4.1	Muon Detection	264
		10.4.2	K_L Detection	266
	10.5	Detect	or R&D	266
		10.5.1	Module Tests and Results	267
			10.5.1.1 Scintillators	267
			10.5.1.2 Fibers	267
			10.5.1.3 Photodetectors	268
			10.5.1.4 Other related studies	269
			10.5.1.5 New R&D studies	269
			10.5.1.6 Radiation Damage Studies	269
			10.5.1.7 New Radiation Damage Studies	269
		10.5.2	Prototype Test and Results	270
		10.5.3	Design and construction of the IFR prototype	270
			10.5.3.1 Beam Tests	271
			10.5.3.2 Tests Results	271
	10.6	Baselir	ne Detector Design	273
		10.6.1	System Layout	273
		10.6.2	Chamber Construction and Assembly	274
	10.7	Front-	End Electronics	274
		10.7.1	Introduction	274
		10.7.2	Photodetectors and PCBs	274
			10.7.2.1 Photodetector PCB and optical coupling to fibers	274

			10.7.2.2	Optical of	couplin	ig to fil	oers .												275
			10.7.2.3	Photode	tector 1	locatio	n												275
			10.7.2.4	Photode	tector o	choice .													276
			10.7.2.5	Aging ar	ıd back	ground	d issue	es											277
			10.7.2.6	Tempera	ture re	equirem	ients												277
		10.7.3	IFR read	lout elect:	ronics:	an ove	erview												278
			10.7.3.1	Introduc	tion .														278
			10.7.3.2	Basic fea	atures o	of the I	FR de	etecto	r										278
			10.7.3.3	IFR chai	nnel co	unt est	imatio	on											279
			10.7.3.4	Estimati	ons of	the IF	R ever	nt size	and	data	bar	ıdwi	idth						279
			10.7.3.5	Backgrou	und rad	diation	and e	lectro	nics o	desig	n co	nstr	aint	. s					280
			10.7.3.6	The IFR	reado	ut syst	em .												283
	10.8	Final a	ssembly a	and instal	llation														291
	10.9	ES&H	issues .																292
	10.10	Structu	re of the	IFR grou	цр.,														292
	10.11	Cost ar	nd schedu	ıle	· · · ·														292
11	Magi	net and	l Flux Re	eturn															297
	0																		
12	Elect	ronics,	Trigger,	Data Ac	quisiti	on and	l Onli	ne											299
	12.1	Archite	ecture Ov	verview .															299
		12.1.1	Trigger S	Strategy															299
		12.1.2	Trigger F	Rate and	Event	Size Es	stimat	ion .											300
		12.1.3	Dead Tir	me and B	uffer Q	Queue I	Depth	Consi	derat	ions									301
	12.2	Trigger	and Eve	ent Data (Chain														302
		12.2.1	Choice o	f Global (Clock I	Frequer	ncy .												302
		12.2.2	Level-17	Frigger .															303
		12.2.3	Fast Con	ntrol and '	Timing	g Syste	m												307
		12.2.4	Control a	and Data	Links														310
		12.2.5	Common	ı Front-Ei	nd Elec	etronics	5												314
		12.2.6	Read-Ou	ıt Module	s														315
		12.2.7	Network	Event Bi	uilder .														316
		12.2.8	High-Lev	vel Trigge	r Farm														317
		12.2.9	Data Log	gging															318
	12.3	System	Integrat	ion and F	Error H	andling	g												318
	12.4	Control	l Systems	3			,												318
		12.4.1	Electroni	ics Contro	ol Syste	em .													319
		12.4.2	Detector	Control	\mathbf{System}														320
		12.4.3	Farm Co	ontrol Svs	tem .														320
	12.5	Other S	Systems																320
		12.5.1	Data Qu	ality Mor	nitoring	r Syste	m												321
		1252	Other Co	omponent	s	,~,									•		•		321
		12.5.3	Software	Infrastru	~		· · · ·			•••					·		•		321
			~01000000	THE OUT U			• • •			•••		• •	• •	• •	•	• •	•	•••	<u> </u>
	-12.6	R&D to	or Electro	onics. Trie	ger an	d Data	a Acar	isitio	n anc	l On	line								321
	$12.6 \\ 12.7$	R&D to Conclu	or Electro sions	onics, Tri _{	gger an	ıd Date	a Acqu	uisitio	n and	l On	line	•••	•••	•••	•	•••	•		321 322

13 Sub	detecto	or Electronics and Infrastructure				325
13.1	Subsys	stem-specific Electronics				325
	13.1.1	SVT Electronics				325
		13.1.1.1 SVT Summary				327
	13.1.2	DCH Electronics				327
		13.1.2.1 Design Goals				327
		13.1.2.2 DCH Front-end system (block diagram)				327
		13.1.2.3 Standard Readout - OFF DETECTOR electronics				328
		13.1.2.4 Sampled Waveforms - OFF DETECTOR electronics				329
		13.1.2.5 Front End Crates				330
		13.1.2.6 Number of crates and links				330
		13.1.2.7 ECS				330
		13.1.2.8 Cabling				330
		13.1.2.9 Power Requirements				330
		13.1.2.10 DCH Summary				331
	13.1.3	PID Electronics				331
		13.1.3.1 The Front-end Crate				331
		13.1.3.2 The Communication Backplane				332
		13.1.3.3 The PMT Backplane				332
		13.1.3.4 Cooling and Power Supply				332
		13.1.3.5 The Front-end Board				332
		13.1.3.6 The Crate Controller Board (FBC)				332
		13.1.3.7 PID Summary				333
	13.1.4	EMC Electronics				333
		13.1.4.1 EMC Summary				334
	13.1.5	IFR Electronics				335
		13.1.5.1 IFR Summary				338
13.2	Electro	onics Infrastructure				338
	13.2.1	Power supplies, grounding and cabling				338
		13.2.1.1 Power Supply to the Front-end:				338
		13.2.1.2 High Voltage Power Supply to the Detectors:				342
	13.2.2	Grounding and Shielding				342
	13.2.3	Cable Plant				342
14 Soft	ware ar	nd Computing				345
14.1	Tools t	to support detector studies	• •		• •	345
	14.1.1	Full Simulation	• •			345
		14.1.1.1 Bruno: the SuperB full simulation software	• •			346
		14.1.1.2 Geometry description	• •			346
		14.1.1.3 Simulation input: Event generators	•••	•••	• •	346
		14.1.1.4 Simulation output: Hits and MonteCarlo Truth	• •		• •	347
		14.1.1.5 Simulation optimization	•••			347
		14.1.1.6 Staged simulation				347
		14.1.1.7 Interplay with fast simulation	• •			348
		14.1.1.8 Long term evolution of the full simulation software	• •			348
	14.1.2	Fast Simulation	• •			349
		$14.1.2.1 \text{Event generation} \dots \dots \dots \dots \dots \dots \dots \dots \dots $				349

		14.1.2.2	Detector description
		14.1.2.3	Interaction of particles with matter
		14.1.2.4	Detector response
		14.1.2.5	Reconstruction
		14.1.2.6	Machine backgrounds
		14.1.2.7	Analysis tools
		14.1.2.8	Simulation validation and detector studies
	14.1.3	Distribut	ted computing tools
		14.1.3.1	Distributed resources
		14.1.3.2	Distributed systems design: a bird's-eye view
		14.1.3.3	The production system
		14.1.3.4	The data analysis system prototype
		14.1.3.5	The bookkeepieng and data placement database
	14.1.4	Collabor	ative tools
		14.1.4.1	Overview
		14.1.4.2	Authorization
		14.1.4.3	Portal System
		14.1.4.4	Document repository
		14.1.4.5	Documentation
		14.1.4.6	Code repository
		14.1.4.7	Code packaging and distribution
14.2	Compu	uting mod	lel outline
	14.2.1	Data pro	ocessing
	14.2.2	Resource	$e estimate \dots \dots$
	14.2.3	Computi	ng Infrastructure
14.3	R & D	program	
	14.3.1	R& D or	parallelization
	14.3.2	GPU R&	z D
	14.3.3	Framewo	ork R & D
	14.3.4	DIRAC d	framework evaluation
		14.3.4.1	Pilot jobs model
		14.3.4.2	Dirac data management
		14.3.4.3	DIRAC API
		14.3.4.4	User Management
		14.3.4.5	Tested Use Cases
		14.3.4.6	SuperBDIRAC module
		14.3.4.7	Building up a DIRAC Infrastructure for SuperB
		14.3.4.8	Future Works
	14.3.5	Data ma	nagement and distributed storage R&D
		14.3.5.1	WAN data access
		14.3.5.2	Data access library
		14.3.5.3	File Transfer Service evolution
		14.3.5.4	Dynamic file catalogue technology
		14.3.5.5	Storage system evaluation
14.4	Outloo	ok	

16 Facilities, Mechanical Integration and Assembly	379
16.1 Introduction	379
16.1.1 Magnet and Instrumented Flux Return	379
16.2 Component Extraction	380
16.3 Component Transport	381
16.4 Detector Assembly	382
17 The SuperB Collaboration and Project Management	383
17.1 Collaboration Membership	383
17.2 The Super B Collaboration Council $\ldots \ldots \ldots$	384
17.3 The SuperB Spokesperson \ldots	385
17.4 The Super B Executive Board \ldots	385
17.5 The Super B Management Team and Management Plan	386
17.6 International Finance Review Committee	386
17.7 Interaction with the Cabibbo-Lab	386
17.8 Communications	386
17.9 Construction Responsibilities	387
18 Cost and Schedule	391
18.1 Detector Costs	392
18.2 Basis of Estimate	396
18.3 Schedule	397

7 Drift Chamber

7.1 Overview - Finocchiaro, Roney 10 pages

7.1.1 Physics Requirements - 3 pages

7.1.2 Geometrical Constraints

The Drift Chamber inner radius is constrained by the final focus cooling system and by the Tungsten shield surrounding it to $R_{inner}^{DCH} =$ 265 mm, the outer radius is constrained to $R_{outer}^{DCH} = 809 mm$ by the DIRC quartz bars. The total length available for the Drift Chamber is of $L^{DCH} = 2700$ mm. As the rest of the detector, the drift chamber is shifted by the nominal *BABAR* offset (367 mm) with respect to the interaction point.

Simulation studies performed on several signal samples with both high $(e.g. B \rightarrow \pi^+\pi^-)$, and medium-low $(e.g. B \rightarrow D^*K)$ momentum tracks indicate that:

a) as expected, momentum resolution improves



Figure 7.1: Track momentum resolution for different values of the drift chamber inner radius.

as the minimum drift chamber radius R_{min} decreases, see Fig. 7.1; R_{min} is actually limited by mechanical integration constraints with the cryostats and the radiation shields.

b) The momentum and especially the dE/dxresolution for tracks going in the forward or backward directions are clearly affected by the change in number of measuring samples when the chamber length is varied by 10-30 cm. However the fraction of such tracks is so small that the overall effect is also not large (to be better quantified with updated FastSim studies).

- 7.1.3 Machine Background Considerations - Cenci 3 pages
- 7.1.4 DCH Design Overview 2 pages
- 7.1.5 Expected Performance 2 pages

7.2 Design Optimization Finocchiaro, Hearty, Piccolo, Roney 9 pages

The BABAR drift chamber operational record has been quite good, both for what performances and reliability is concerned; however there are possible paths of improvements that we have explored to try and design a drift chamber with performances yet better that the BABAR one.

7.2.1 Cluster Counting

A possibility being considered to improve the performances of the gas tracker is the use of the cluster counting method. Signals in drift chambers are usually split to an analog chain which integrates the charge, and to a digital chain recording the arrival time of the first electron, discriminated with a given threshold. The *clus*ter counting technique consists instead in digitizing the full waveform to count and measure the time of all individual peaks. On the assumption that these peaks can be associated to the primary ionization acts along the track, the energy loss and to some extent the spatial coordinate measurements can be substantially improved. In counting the individual cluster, one indeed removes the sensitivity of the specific energy loss measurement to fluctuations in the amplification gain and in the number of electrons produced in each cluster, fluctuations which significantly limit the intrinsic resolution of conventional dE/dx measurements.

The ability to count the individual ionization clusters and measure their drift times strongly depends on the average time separation between them, which is, in general, relatively large in He-based gas mixtures thanks to their low primary yield and slow drift velocity. Other requirements for efficient cluster conting include good signal-to noise ratio but no or limited gasgain saturation, high preamplifier bandwidth, and digitization of the signal with a sampling speed of the order of 1Gs/sec. Finally, it is necessary to extract online the relevant signal features (*i.e.* the cluster times), because the DAQ system of the experiment would hardly be able to manage the enormous amount of data from the digitized waveforms of the about 10 000 drift chamber channels.

7.2.2 Cell Design and Layer Arrangement

The drift chamber cell design must optimise the homogeneity of the electric field inside the cell; this is particularly relevant with the nonsaturated mixture we intend to use. Another critical parameter is the overall wire material, and an optimal use of the drift chamber volume for accommodating as many measurament points as possible.

The design for the Super*B* drift chamber employs small rectangular cells arranged in concentric layers about the axis of the chamber. The *z* coordinate of the track hits is measured by orienting a subset of the wire layers at a small positive or negative stereo angle, ε , relative to the chamber axis. Such a measurement is performed with precision $\sigma_z \simeq \sigma_{R\phi}/\tan\varepsilon$. As in *BABAR*, four consecutive cell layers are grouped radially into a superlayer (SL). This will allow to keep the same *BABAR* algorithms for track-segment finding, both in the track reconstruction and in the formation of the drift chamber trigger.

The rectangular cell layout ensures the most efficient filling of the drift chamber volume, because the transition between superlayers of opposite stereo angles does not require to leave free radial space, nor layers of field-shaping guard wires. Indeed, the latter are only used at a radius inside the innermost SL and at a radius outside the outermost SL. Such guard wires also serve the purpose to electrostatically contain very low momentum electrons produced from background particles showering in the DCH inner cylinder and in the SVT, or backgroundrelated backsplash from detector material just beyond the outer SL.

Simulations[1] have shown that a field:sense wire ratio of 3:1 ensures good homogeneity of the electric field inside the cells. In this configuration each sense wire is surrounded by 8 field wires.

The radial positions of the stereo wires in the *j*-th layer vary with the *z* coordinate, being larger at the endplates than at the center of the chamber by the "stereo drop" $\delta_j \equiv R_j^{\rm EP} - R_j$. The cell shapes are most uniform when $\delta_j = \delta$ is a constant for all layers: this is obtained by changing the stereo angle with the radius, by the relation $\tan \varepsilon_j = 2\delta/L_j\sqrt{2R_j^{\rm EP}/\delta - 1}$ (L_j is the chamber length at layer *j*).

Additional constraints used to determine the cell layout include:

- a) the number of cells of width w_j on the *j*-th sense wire layer, $N_j = 2\pi R_j/w_j$, must be an integer number;
- b) to keep a fixed periodicity in signal and high voltage distribution, it is convenient that the number of cells per layer is incremented of a fixed quantity ΔN when passing from a SL to the next one.
- c) since the density of both physical tracks and background hits is higher at smaller radii, we choose to have smaller cells in the innermost layers of the drift chamber.

A possible choice for the drift chamber layout, obtained for k = 4, $\delta = 8 \text{ mm } \Delta N = 16$ is shown in Table 7.1. In this arrangement the two innermost SL's contain 1472 cells with height $h = 10 \,\mathrm{mm}$ and widths $w = (10.2 \div 11.7) \,\mathrm{mm}$. The cells in the remaining superlayers have h =13 mm and $w = (16.1 \div 19.1)$ mm. There are a total of 7872 cells in the drift chamber. The first two superlayers have an axial orientation; this minimizes the occupancy from background hits due to low-momentum spiraling electrons which traverse the drift chamber along its axis (see Sec.7.1.3). The two external super layers are also axial. The fact that the innermost and outermost super layers do not exhibit the stereo drop deformation δ matches the axial simmetry of the inner and outer drift chamber cylinders. The six internal SL's have a stereo arrangement, with angles as shown in the Table.

Table 7.1: A possible drift chamber SL structure, specifying the number of cells per layer, the radius at the center of the chamber of the innermost sense wire layer in the SL, the cell widths, and wire stereo angles, which vary over the four layers in a SL as indicated.

SL	N _{cells}	R	width	Angle	
		[mm]	[mm]	[mrad]	
1	176	286.0	10.2 - 11.3	0	
2	192	326.0	10.7 - 11.7	0	
3	144	369.0	16.1 - 17.8	+(63-66)	
4	160	421.0	16.5 - 18.1	-(67 - 69)	
5	176	473.0	16.9 - 18.3	+(70-72)	
6	192	525.0	17.2 - 18.5	-(73 - 75)	
7	208	577.0	17.4 - 18.6	+(76-77)	
8	224	629.0	17.6 - 18.7	-(78-79)	
9	240	689.0	18.0 - 19.1	0	
10	256	741.0	18.2 - 19.1	0	

The corresponding wire map in the region with angle $|\varphi| < 10^{\circ}$ is shown in Fig. 7.2 at the center of the chamber (a) and at the endplates (b).

It is seen that the axial-stereo transition between SL 2 and SL 3 creates some additional radial space close to the endplates, which disappears at the DCH center. The opposite happens at the stereo-axial transition between SL 8 and SL 9. It is clear that the electric field should be as uniform as possible across layers to ease the drift chamber calibration; however, simulation studies have shown that the field distortion at the two transition radii is moderate and does not require to be compensated by layers of guard wires, which would add material and reduce the sensitive volume.

In Fig.7.3 we show the drift lines and the isochrone curves for two sample rectangular cells of the proposed Super*B* drift chamber in a 90%He- $10\%iC_4$ H₁₀, in a 1.5 T magnetic field. The rectangular cells with field:sense wire ratio of 3:1 are indeed a satisfactory compromise, ensuring that the field lines are sufficiently con-

tained within the cell and the isochrone lines are isotropic for most of the drift region, while at the same time the number of field wires is not excessively large.

7.2.3 Gas Mixture

The gas mixture for the SuperB drift chamber is chosen to allow optimal resolution in the measurement of both momentum and energy loss. It must also be operationally stable (e.g.,have a wide high voltage plateau), and be little sensitive to photons with $E \leq 10 \,\mathrm{keV}$ to help controlling the rate of background hits (see Sec. 7.1.3). Finally, aging in the chamber should be slow enough to match the projected lifetime of a typical High Energy Physics experiment (about 15 years). These requirements already concurred to the definition of the BABAR drift chamber gas mixture, (80%He-20%iC₄H₁₀). Indeed, a high Helium content reduces the gas density and thus the multiple scattering contribution to the momentum resolution. Good spatial resolution calls for high single electron efficiency and for small diffusion coefficient. The effective drift velocity in Helium-based gas mixtures is typically non saturated, therefore it depends on the local electric field, and on the Lorentz angle. This dependences can be taken into account by a proper calibration of the space-time relations and in principle do not pose limits to attaining the required spatial resolution. In practice, a careful choice of the cell shape (see the discussion in Sec. 7.2.2), and a small value of the Lorentz angle are an advantage.

To match the more stringent requirements on occupancy rates of Super*B*, it could be useful to select a gas mixture with a larger drift velocity in order to reduce ion collection times and so the probability of hits overlapping from unrelated events. The *cluster counting* option would instead call for a gas with low drift velocity and primary ionization. As detailed in Section7.2.4, R&D work is ongoing to optimize the gas mixture.

7.2.4 R&D and Prototype Studies

In order to optimize the gas mixture for the SuperB environment, and to asses both the feasibility and the operational improvements for the *cluster counting* technique a complete R&D



Figure 7.2: A possible cell layout of the Super*B* drift chamber with $h_{in} = 10 \text{ mm}$, $h_{out} = 13 \text{ mm}$. Open green squares: guard wires; open blue circles: field wires; full red circles: sense wires. Note how the boundary regions after the first 8 layers of axially strung wires in the inner part of the chamber and after the following 24 layers of stereo layers map differently at the drift chamber center and at the endplates.



(a) Field and isochrone lines in a sample "smal" cell, on layer n. 6.



(b) Field and isochrone lines in a sample "big" cell, on layer n. 22.

Figure 7.3: Field lines and isochrone curves (shown with a 20 ns step) in a cell belonging to the first 8 layers (left) and in a larger cell of the outermost 32 layers (right).

layer	N_{cells}	R	width	Angle	la	yer	N_{cells}	R	width	Angle
		[mm]	[mm]	[mrad]				[mm]	[mm]	[mrad]
1	176	286.0	10.2	0.0		21	192	525.0	17.2	-72.9
2	176	296.0	10.6	0.0		22	192	538.0	17.6	-73.5
3	176	306.0	10.9	0.0		23	192	551.0	18.0	-74.2
4	176	316.0	11.3	0.0		24	192	564.0	18.5	-74.9
5	192	326.0	10.7	0.0		25	208	577.0	17.4	75.5
6	192	336.0	11.0	0.0		26	208	590.0	17.8	76.1
7	192	346.0	11.3	0.0		27	208	603.0	18.2	76.7
8	192	356.0	11.7	0.0		28	208	616.0	18.6	77.3
9	144	369.0	16.1	62.9		29	224	629.0	17.6	-77.8
10	144	382.0	16.7	63.9		30	224	642.0	18.0	-78.4
11	144	395.0	17.2	64.8		31	224	655.0	18.4	-78.9
12	144	408.0	17.8	65.7		32	224	668.0	18.7	-79.4
13	160	421.0	16.5	-66.6		33	240	689.0	18.0	0.0
14	160	434.0	17.0	-67.5		34	240	702.0	18.4	0.0
15	160	447.0	17.6	-68.3		35	240	715.0	18.7	0.0
16	160	460.0	18.1	-69.1		36	240	728.0	19.1	0.0
17	176	473.0	16.9	69.9		37	256	741.0	18.2	0.0
18	176	486.0	17.4	70.7		38	256	754.0	18.5	0.0
19	176	499.0	17.8	71.4		39	256	767.0	18.8	0.0
20	176	512.0	18.3	72.2		40	256	780.0	191	0.0

Table 7.2: A possible drift chamber layer structure, specifying the number of cells per layer, the wire layer radius at the center of the chamber, the cell width and the wire stereo angle.

program has been proposed. The program includes both beam tests and cosmic ray stands to monitor performances of *ad hoc* built prototypes. While the dE/dx resolution gain of the cluster counting method is in principle quite sizeable compared to the traditional total charge collection, the actual capability of the measured number of cluster might not retain the same analyzing power, due to a pletora of experimental effects that should be studied in detail so that the energy loss measurement derating should be assessed and, if possible, cured.

A few prototypes were built and operated to answer the above mentioned questions.

7.2.4.1 Prototype 1

The first one is a small aluminum chamber, 40 cm long, with a geometry resembling the the original *BABAR* drift chamber. It consists of 24 hexagonal cells organized in six layers with four cells each. A frame of guard wires with appropriate high voltage settings surrounds around the cell array to ensure uniformity of the electric field among the cells. The device was operated in a cosmic ray test stand in conjunction with an external telescope, used to extapolate the track trajectories with a precision of $80 \,\mu\text{m}$ or better . Different gas mixtures have been tried in the prototype: starting with the original *BABAR* mixture (80%He-20%iC₄H₁₀) used as a calibration point, both different quencher proportions and different quenchers have been tested in order to assess the viability of lighter and possibly faster operating gases.

As an example, the correlation between the extrapolated drift distance and the measured drift time is shown in Fig. 7.4 for a 75%He-25%C₂H₆ gas mixture. The result of a fit to a 5^{th} -order Chebychev polynomial is superimposed to the experimental points. Track-fit



Figure 7.4: Track distance vs. drift time in a cell of the prototype. The line is the result of a fit with a 5^{th} -order Cheby-chev polynomial.

residuals and spatial resolution as a function of the drift distance for the same gas mixture are show in Fig. 7.5.



Figure 7.5: Track fit residuals (top) and spatial resolution (bottom) as a function of the drift distance.

7.2.4.2 Prototype 2

A full-length drift chamber prototype was designed, built and commissioned to study *cluster counting* in a realistic environment, including signal distortion and attenuation along 2.5 meter long wires. The prototype, which is also meant to serve as a test bench for the final Front-End electronics and for the drift chamber trigger, is composed by 28 square cells with 1.4 cm side, arranged in eight layers and – as in the final SuperB drift chamber – with a field-to-sense wire ratio of 3:1. The eight layers have either 3 or 4 cells each, and are staggered by half a cell side to help reduce the left-right ambiguity. Tracks with angle $|\vartheta| \leq \pm 20^\circ$ cross all the eight layers of the chamber. A set of guard wires surrounds the matrix of 28 cells to obtain a well-behaved field distribution at the boundary of the active detector volume. Most of the cells feature a $25 \,\mu \text{m}$



Figure 7.6: Prototype 2: detail of the strung wires.

Gold-plated Molybdenum sense wire, while for reference seven cells in two adjacent layers are strung with a $25 \,\mu$ m Gold-plated Tungsten wire, traditionally used in drift chambers. The reason for using the Molybdenum wire is its lower resistivity, therefore smaller dispersion for pulses

travelling along the wires. A picture of the chamber after stringing completion is shown in Fig. 7.6. The entire wire structure is enclosed in



Figure 7.7: Prototype 2: FEE Faraday cage with signal and trigger cables.

an Aluminum container 3 mm thick; three pairs of thin windows have been carved in the middle and at the extremities in order to have smaller amount of material in the path of low energy particles measured by the device. Four preamplifier boards are used to extract the cell signals. Each board serves seven channels, each with a transimpedence preamplifier (rise time of about 2.4ns), at a nominal gain of 8mV/fC and a noise of 2200 e rms. Each boards also has a test input, both unipolar and differential outputs $(50 \Omega - 110 \Omega)$; the latter are used for a test implementation of the Drift Chamber first level trigger. A detail of the Faraday cage housing the FEE boards and of the signal and trigger cables is shown in Fig. 7.7.

The data collected with this prototype are fed into a switch capacitor array digitizer¹, which samples the wire signals at 1 GS/sec with and input BW > 500 MHz. The challenge of detecting the ionization clusters in signals with a wide dynamic range and non-zero noise levels is apparent from the two sample waveforms shown in Figg. 7.8, recorded in the cosmic-ray setup. Hits associated to cosmic ray tracks reconstructed in the drift chamber prototype are used to compare the performances in the energy loss measurement of the traditional truncated mean algorithm and of the *cluster counting* method. Preliminary results when 10 samples from a single prototype cell are used to form a 70% truncated mean or to count the average number of clusters are shown in Fig. 7.9. In the experimental conditions of our test, *cluster counting* yields a 40-50% better relative resolution than the truncated mean method. Additional R&D efforts are ongoing to extend this encouraging result to different momentum regions, and study how the $K - \pi$ resolving power in the range of interest of Super $B(|p| \leq 5 \,\text{GeV}/c)$ improves with the *clus*ter counting technique.

7.2.4.3 Single Cell Prototypes

Within the past year, the group has undertaken two beam tests of single-cell drift chamber prototypes at the TRIUMF M11 beam line. The goals of the tests were:

- to establish the benefits of clusters counting for particle identification;
- to study the suitability for cluster counting of various amplifier prototypes provided by the University of Montreal; and
- to quantify the impact on particle identification performance of various design choices, including sense wire diameter, cable for transmitting the analog signal, connectors, termination, and gas gain.

The beam test in November 2011 used a single prototype, with $25 \,\mu\text{m}$ diameter sense wire, while the test in summer 2012 used two prototypes, one with $20 \,\mu\text{m}$ sense wire, and the other with either 25 or $30 \,\mu\text{m}$. The prototypes were 2.7m long, and consisted of a single 15 mm square cell surrounded by an array of

¹CAEN V1742: http://www.caen.it/csite/CaenProfList. jsp?parent=13&Type=WOCateg



Figure 7.8: Sample waveforms from two cells of the full-length drift chamber prototype.

bias wires that adjusted the electric field distribution within the cell to be that expected for a large drift chamber [Fig. 7.10]. A 90% helium / 10% isobutane (90:10) mixture was used for the summer 2012 tests, while the November 2011 test also tested 80:20 and 95:5.

J. P. Martin (U. Montreal) provided five different amplifier prototypes, with three different input impedances: 50, 170, and 380 Ω . The impedance of the cell is 380 Ω , and for most tests, the cell was terminated at this value on the non-readout end. The 170 and 380 Ω consisted of an impedance-matching front end, followed by a 100× gain stage (of 50 Ω input impedance); the 50 Ω amplifier consisted only of this gain stage.

The TRIUMF M11 beam line was used for the tests [Fig. 7.11a]. Our tests used positivelycharged electrons, muons, and pions, with momenta ranging from 140 to 350 MeV/c. In this momentum range, the particle identification separation between muons and pions is comparable to that between pions and kaons at the 2-3 GeV/c range relevant for Super*B*. The trigger and time-of-flight (TOF) system consisted of two scintillator counters, each with a pair of Burle micro-channel plate phototubes. The trigger rate was typically tens of Hertz, and the TOF resolution was 130 ps, providing clean separation between the particle species.

In the November 2011 test, the drift chamber waveforms were digitized using a CAEN switched-capacitor array, read out using the MI-DAS data acquisition system. The bandwidth of this module is 300 MHz, which may be less than required for cluster counting, so in the Summer of 2012, a 4 GHz bandwidth LeCroy oscilloscope was instead used for digitization. In both cases, the amplifier and digitizer were connected by a



Figure 7.9: Average dE/dx and number of clusters from 10 samples of a single cell belonging to a track reconstructed in the prototype.

10 m long cable, the distance expected in the final design. Events were written to disk at 10–15 Hz. Figure 7.12 shows a typical waveform.

The benefits of cluster counting on particle ID performance are characterized by comparing the separation between muon and pion tracks using dE/dx only to the combination of dE/dx and cluster counting. Analysis is in progress, and results shown here are preliminary.

A track is formed by randomly selecting 40 samples of the same particle species as determined by the TOF system. The track dE/dx is obtained by discarding the largest 30% of the samples. Clusters are identified by a simple threshold on a smoothed version of the waveform. This algorithm is not necessarily optimal, and other methods are under study. Conversely, adequate performance may be achieved using a

simpler algorithm that could be implemented in hardware, as opposed to an FPGA. The track is characterized by the average of the number of clusters in the 40 samples. The track dE/dxand cluster counting values are combined in a likelihood ratio that is used to label the track as a muon or pion. Figure 7.13 show examples of the results from the two beam tests. The addition of cluster counting significantly reduces the fraction of muons that are misidentified as pions.

7.2.4.4 Aging studies

The goal of the aging studies is to establish that the proposed drift chamber can survive for a lifetime of at least 100 ab^{-1} .

These studies use an 55 Fe source both to age a test chamber, and to characterize its



Figure 7.10: Cell design of the single cell prototypes.



Figure 7.11: The M11 beam area at TRIUMF, showing two single-cell drift chamber prototypes under test in Summer 2012.

performance. The initial studies are using a 30 cm long chamber containing a single *BABAR*-like hexagonal-shaped cell, and an 80% helium 20% isobutane gas mixture. The sense wire is 20 micron gold-coated tungsten, and the field and bias wires are 120 micron gold-coated aluminum. The chamber is exposed to a 100 mCi ⁵⁵Fe source. The resulting current is monitored, along with temperature and atmospheric pressure, as a way to characterize gain as a func-





Figure 7.12: Waveforms recorded in Summer 2012 from a prototype with $20 \,\mu\text{m}$ sense wire (red) and $30 \,\mu\text{m}$ sense wire (blue), using two different $370 \,\Omega$ input impedance amplifier prototypes. The yellow curve shows the NIM logic signals from the time-of-flight system.

tion of accumulated charge. Once per week, the hot source is replaced with a low-intensity one and the pulse-height spectrum is recorded (Fig. 7.14). The location of the 55 Fe peak is an additional measurement of gain. The number of very small pulses is sensitive to the Malter effect, a form of aging of the field wires in which they accumulate an insulating coating. A second single-cell chamber, which is not exposed to the hot source, is used to calibrate out any possible gain effects due to gas variations, and to verify the gas density corrections. The chamber is operated at a voltage such that the electric field on the field wires is less than 20 kV/cm in order to minimize the Malter effect.

The aging chamber shows a gain drop of 25% after accumulating 310 mC/cm over the last 20 months. This lifetime is significantly in excess of the 34 mC/cm accumulated by the *BABAR* drift chamber. The *BABAR* chamber saw a 10% loss of gain over that time.

The next aging chamber is currently under construction. It will include seven square cells, so that the central field wires are surrounded by sense wires, and will therefore accumulate the correct amount of charge (Fig. 7.15). It will use Super*B* materials, gold-coated 20 micron Molybdenum sense wires, and 80 or 90 micron bare aluminum field wires, and 90:10 helium-



Figure 7.14: Pulse height spectrum recorded from an ⁵⁵Fe source by an aging chamber. The red curve is the underlying cosmic ray background.

isobutane gas. The structure of the chamber will be aluminum, as for the current aging chamber, but the walls will be covered by samples of the carbon-fiber material that will be used in the actual chamber.



Figure 7.15: Wire layout of the new aging chamber under construction.

The amount of charge per cm expected for the SuperB chamber is a function of the chamber occupancy, the gas gain, and the total running time. Current background calculations indicate occupancy levels comparable to BABAR,



Figure 7.13: Fraction of muons identified as pions versus the fraction of pions satisfying the selection criteria, for dE/dx only, and for the combination of dE/dx and cluster counting. (a) 140 MeV/c data, November 2011 test; (b) 210 MeV/c data, summer 2012 test.

including a five-times safety factor. The SuperB running time will also be comparable to BABAR. The gas gain may be higher, due to the requirements of cluster counting. The gas gain required is a function of the amplifier; the 50 ohm input impedance amplifier prototypes require four times the gas gain of the 370 ohm amplifiers, due to the impedance mismatch between the amplifier and the drift cell. The field wire diameter is, in turn, a function of the sense wire operating voltage, given the need to keep the electric field at the surface of the field wire below 20 kV/cm.

7.2.5 R&D Future Developments

- test beam with different particle species

- on-board feature extraction
- AOB

7.3 Mechanical Design

The drift chamber mechanical structure must sustain the wire load with small deformations, while at the same time minimizing material for the surrounding detectors. The structure is also required to ensure tightness for the gas filling the drift volume. We opted for a structure entirely in Carbon Fiber (CF) composite, with an approximately cylindrical geometry. A side view of the chamber is shown in Fig. 7.16.



Figure 7.16: Longitudinal section of the DCH with principal dimensions.

7.3.1 Endplates

The wires defining the cell layout are strung between the two endplates, which are required to:

- a) sustain the total wire load of XXXX tons (or N) (see sec. ??) with minimal deformations;
- b) be as transparent as possible to avoid degrading the performances of the forward calorimeter.
- c) have XXXXX precisely machined holes to allow positioning the crimp feed-throughs with tolerances better than XXXX μ m;

The endplates are two identical pieces of 8 mm thick CF composite with inner radius of 250 mm and outer radius of 805 mm. Deformations under load can be minimized using for the endplates a shaped profile. An optimization taking into account different constraints resulted in spherical convex endplate, with a radius of curvature of 2100 mm. Two CF stiffening rings on the inner and outer rims help preventing radial (axial) deformations. An intermediate modulus carbon fiber (as T300 (XXXX), with a Young modulus of XXXXX) will be used. It is expected that the average material characteristics will be degraded by about 30 % or less after drilling the XXXX holes on the endplates. Detailed studies on this aspect will be performed on custom samples. The maximum displacements on the endplates is calculated to be less than $300\,\mu\text{m}$ (Figure ??).

7.3.2 Inner cylinder

The drift chamber inner cylinder should be as transparent as possible to minimize the multiple scattering degradation to the p_T measurement. For this reason it was designed as a non load-bearing structure: it must only guarantee gas tightness, and sustain possible differential pressures of the order o 10 mbar between the inside and outside of the chamber. It is a thin (200 μ m) CF cylinder of 250 mm radius, with a 25 μ m aluminum foil glued on it as RF shield. During the stringing phase the inner cylinder will be free to move longitudinally being fixed only to one



Figure 7.17: Displacement of each endplate due to the wire load.

endplate. Only after stringing, when all endplate deformations are settled, it will be glued to the other endplate.

7.3.3 Outer Cylinder

In addition to guaranteeing gas tightness and with standing a differential pressure as the inner cylinder, the outer cylinder will also carry the wire load. It will be installed after completion of the wire stringing. To ease the construction and the mounting procedures, the cylinder is longitudinally divided in two half shells. Each shell consists of two 1 mm-thick CF skins laminated on a 6 mm-thick honeycomb core. Two thin aluminum foils, $100 \,\mu$ m on inside surface and $25 \,\mu$ m on outside surface, are glued to the shells to ensure the rf shield. The sandwich structure guarantees a high bending stiffness and a high safety factor for global buckling.

7.3.4 Choice of wire and electrostatic stability

7.3.5 Feedthrough design

The feedthroughs locate the wires to within the specified tolerances, hold the wire tension, and, in the case of the sense wires, insulate against the high voltage. They must achieve these goals while maintaining a helium-tight gas seal.

A feedthrough is made from two components, a plastic outer insulator, and a conducting crimp pin (Fig. 7.18). The insulators are injectionmolded parts formed from Celenex 3300-2, chosen for its low shrinkage during molding, dimensional stability, and high dielectric strength. The crimp pins for the aluminum field wires are aluminum 6063. Studies are planned to determine whether copper or aluminum crimp pins are more suitable for the molybdenum sense wires. The crimp pins will have a gold-flash coating with a nickel underlayment.



Figure 7.18: Sense wire (top) and field wire (bottom) feedthroughs from the *BABAR* drift chamber. The SuperB parts will be similar.

The two parts are glued together with an epoxy that is dyed so that extraneous epoxy on the crimp pin can be identified and removed.

The tolerances are specified to ensure that the contribution to cell resolution is small, with tolerances on the sense wire parts significantly tighter than those on the field wires. Inner diameter of sense wire crimp pins at the wire release point will be 100 microns, and 200 microns for field wires. Concentricity of the pin hole with respect to the shaft diameter will be less than 30 microns, and eccentricity of the shaft will be less than 25 microns.

Each of the approximately 75000 feedthroughs will be individually tested against the specifications. Sense wire feedthroughs will have an additional test to verify HV performance.

7.3.6 Endplate systems

7.3.6.1 Electronics enclosures

The amplifiers mounted on the backward endplate of the drift chamber, and the high-voltage components mounted on the forward endplate are covered by the electronics enclosures. The volumes are filled with nitrogen to ensure that a leak of the flammable drift gas through the endplate cannot form an explosive mixture. The enclosures also protect the components inside, and provide the mounting points for the chamber as a whole.

The enclosures are light aluminum structures (Fig. 7.19). The main features are feedthroughs for the various signals and services used by the chamber: chamber gas, nitrogen, cooling water, amplifier power and signal, and high voltage. Each 1/16th sector corresponds to approximately 500 signal cables, arranged into 8channel ribbon cables and feedthroughs. These feedthroughs may also carry the power and control lines for each amplifier. The feedthroughs are mounted on removable panels, while the cooling lines are on the fixed ribs. The panels allow access for installation and repairs, although such accesses are expected to be rare. Although the endplates are curved, the panels can be flat, reflecting the geometries of the backwards calorimeter and the forward time-of-flight system on either side of the drift chamber. This will greatly simplify the necessary gas seals.

7.3.6.2 Cooling

The fast amplifiers mounted on the drift chamber endplates produce approximately 1200 W of heat. This heat must be removed to keep the temperature of the drift chamber and nearby detectors stable and uniform. The cooling system to accomplish this will be water based, operating at a pressure below atmospheric pressure. Small leaks will therefore cause air to leak into the water, rather than leading to water leaking into the electronics enclosure. The system is quite similar to one recently built for the near detector of the T2K experiment.

The major components of the system include two water reservoirs, one at atmospheric pres-



Figure 7.19: (a) The rib structure of each enclosure is machined from a solid piece of aluminum. (b) The feedthroughs are on panels mounted to the ribs.

sure, and the other maintained at an absolute pressure of 0.3 atmospheres; a pump to move water between the systems; cooling lines to the drift chamber and inside the electronics enclosures; valves and gauges; a heat exchanger that connects to the laboratory chilled water system; and a control system to maintain the desired water temperature. Appropriate corrosion inhibitors and microbiocides will be added to the water. The lines within the enclosures will be mounted to the fixed ribs, and may include fins or other features to increase the surface area. These features, combined with turbulent nitrogen flow in the enclosure, may generate sufficient thermal flow from the amplifiers to keep the temperatures at an acceptable level. Mockups and thermal calculations will be undertaken to test this concept. The alternative will be cooling straps.

Although there is no heat generated on the forward endplate, cooling lines will be run to ensure a uniform temperature across the drift chamber.

7.3.6.3 Shielding

The aluminum structure of the electronics enclosures, together with the aluminum skins on the outer and inner cylinders, form a Faraday cage that encloses the amplifiers and the chamber wires. A 25 micron thick aluminum skin bonded to the endplates provides additional shielding between the chamber wires and the enclosure volumes.

7.3.6.4 Electromechanical boards

Electrical connections to the crimp pins are required at both ends of every wire. At the forward end, the HV distribution boards ground the field wires and provide HV to the sense wires via the circuit described in Sec. 7.5, and terminate the sense wire at the characteristic impedance of the cell. At the backwards end, service boards ground the field wires, and gather the signals from eight sense wires onto a single multi-conductor connector. The only active components on the service boards are the HV blocking capacitors.

Each of the ten superlayers requires a different size of HV distribution board and service board. Only a single style of 8-channel preamp card is required. Each board will service eight cells (four in radius by two in azimuth), typically corresponding to 32 crimp pins. A number of crimp pin connections much larger than this would make the board difficult to insert and remove.

The connections to the crimp pins are made using low-insertion-force connectors, such as the Hypertronics connectors used for *BABAR*. The details of how the curvature of the endplate is handled in the design of the cards will require prototypes and mockups.

Jumpers between adjacent boards will connect the ground planes. At the HV end, jumpers will distribute the HV among the typically three boards serviced by a single HV channel.

The endplate will include two blind holes per board that will be used with dowel pins to align the boards during insertion. We do not anticipate using pull-down screws, which would require a large number of tapped holes in the endplates. Each board will have two threaded holes that will be used to push against the endplate if it is necessary to remove the board. The boards will be removed rarely, if ever. One of the two holes will be used during normal operations to provide an electrical connection to the aluminum RF shield on the endplate via a lowforce spring connection.

7.3.7 Stringing

7.4 Electronics - Felici, Martin 1 page

7.4.1 Design Goals

The Super*B* Drift Chamber (DCH) front-end electronics is designed to extract and process approximately 8000 sense wire signals to:

- measure the electrons drift times to characterize a track (momentum of charged particles)
- measure the energy loss of particles per unit length, dE/dx (particle identification)
- provide hit information to the trigger system (trigger primitives)

Two scenarios will be investigated for the energy loss measurements. The first one is based on the measurement of the integrated charge on each sense wire, discarding the largest charge values to remove the Landau tails (Standard Readout), while the second one is based on the counting of the primary electron clusters (Sampled Waveforms). Because the front-end requirements for the two options are quite different, each option will be discussed in a separate section.

7.4.2 Standard Readout - charge measurements specifications

The method is based on integrated charge measurement thus allowing the use of (relatively) low bandwidth preamplifiers. This makes the front-end chain less sensitive to noise pickup and instabilities, a suitable condition in a system with a large number of channels.

The three main specifications for a charge measurement are: resolution, dynamic range and linearity.

7.4.2.1 Resolution

Charge measurements for particle identification aim to measure the particle most probable energy loss, to a precision of the order of 7.5% despite the large fluctuations present in a single measurement.

The goal can be achieved by sampling many times the collected charge and applying the "truncated mean" method to resolve the distribution peak value to a precision of several percent.

Because the Super*B* DCH design parameters and expected working conditions aim for an overall single cell resolution (σ_E) of about 35% and σ_E is mainly driven by the detector contribution, we can set a limit of 15% of the resolution for the front-end electronics contribution (i.e. $\sigma_{EL} \sim 5\%$) which can be neglected since $\sqrt{\sigma_E^2 + \sigma_{EL}^2} \sim \sigma_E$.

Finally, if we assume that the charge collection due to a minimum ionizing particle crossing orthogonally the cell is about 50 fC (~ 2fC/e @ 10^5 nominal gas gain) we can infer a limit to the Equivalent Noise Charge (ENC) for a single front-end channel of about 50 $fC \cdot 0.05 \simeq$ 2.5 fC.

7.4.2.2 Dynamic range

With 8 bits ADCs the dynamic is $2.5 - 500 \ fC$. This is more than necessary to satisfy the system requirements.

7.4.2.3 Linearity

As stated above, a single cell energy resolution is about 35%. Therfore a linearity of the order of 2% largely satisfies the system requirements.

7.4.3 Standard Readout - time measurements specifications

As for charge measurements, we have three main specifications: resolution, dynamic range and linearity.

7.4.3.1 Resolution

One of the Super*B* DCH requirements is the charged particle tracks reconstruction. The measurement consist of recording the arrival time at the sense wire of the first ionized electron. This is done with a high precision compared to the spatial resolution (σ_S) of ~ 110 μm . Contribution to σ_S are due to primary ionization statistics, electrons diffusion times and time measurement accuracy.

Assuming an intrinsic chamber resolution (σ_{SC}) of about 100 μm (ref DCH) the upper limit for the electronic contribution can be deduced to be $\sigma_{EL} \leq \sqrt{\sigma_S^2 - \sigma_{SC}^2} \simeq 50 \ \mu m$. As helium based gas mixtures are characterized by a non saturated drift velocity up to high fields, [5] this velocity rapidly increases as the electrons approach the sense wire. A value of 2.5 $cm/\mu s$ (25 $\mu m/ns$) [6] has been used to evaluate the maximum acceptable error in a time measurement, that is $\sigma_t \leq 50[\mu m]/25[\mu m/ns] \simeq 2 \ ns$.

Discarding the bunch length contribution (tenths of ps) there are two main error sources in time measurements: the discriminator jitter and the TDC resolution (digitization noise). The discriminator jitter, in turn, has two main contributions: signal noise and time-walk.

The signal noise contribution is generally small and can be evaluated according to $\Delta t = \sigma_{noise}/(dV/dt) \simeq \sigma_{noise} \cdot \tau/V_{max}$ where τ is the preamplifier-shaper peaking time. Assuming that a single electron cluster generates a signal of amplitude ~ 20 mV, and that the noise and the peaking time associated with the signal are $\sigma_{noise} \sim 3 mV(rms)$ and $\tau \sim 5 ns$ we get a noise contribution to the time resolution of about 0.8 ns.

The time-walk effect is caused by the signal amplitude variation. With a peaking time of about 5 ns, a time-walk contribution for a lowthreshold leading-edge discriminator can be estimated to be about 1.5 ns.

Finally, the digitization noise depends on the digitization unit Δ according to the formula $\sigma = \Delta/\sqrt{(12)}$. Using $\Delta \simeq 1.5 \ ns$ a digitizing noise of about 0.45 ns is obtained.

In summary, without corrections, the time resolution is dominated by time-slewing effects thereby can be estimated to be about 1.8 *ns* (including all contributions). Nevertheless corrections can be applied using digitized signals to minimize time-slewing effects then reducing the time walk contribution (Fig. ??).

7.4.3.2 Dynamic Range

The TDC range depends on the drift velocity and on the cell size. A maximum drift time of about 600 ns has been estimated for Super*B* DCH cells. Providing some safety factor, a TDC range of about 1 μs is enough to include any jitter in trigger generation and distribution.

7.4.3.3 Linearity

A linearity of the order of 1% fully satisfies time measurement requirements.

7.4.4 Standard Readout - DCH Front End Electronics (overall design)

The DCH FEE chain (Fig. ??) is split in two blocks: ON DETECTOR and OFF DETECTOR electronics.

In the following paragraphs we will give a description of the ON DETECTOR electronics while the description of the Off Detector Electronics can be found in Sec. 13.1.2.

7.4.5 Standard Readout - On Detector Electronics

7.4.5.1 Preamplifier Boards

Preamplifier boards will contain HV blocking capacitors, protection networks, preamplifiers and (eventually) shapers-amplifiers. Because of the small cell dimensions many cells can be grouped in a single, multi-channel preamplifiershaper board. Signals and power supply cables will be connected to the boards by means of suitable connectors.

In addition to the requirements on the Signal to Noise Ratio (SNR), each preamplifier should be characterized by enough bandwidth to preserve signal time information and have low power requirement, not more than $20 \div 30 \ mW$ per channel, to limit the total power dissipation on the backward end-plate to $160 \div 240 \ W$, thus allowing the use of simpler and safer forced air based cooling system (no risk of leak).

Concerning the circuit implementation, since

Table 7.3: Preamplifier main specifications

Linearity	< 2%(1 - 100 fC)
Output Signal Umbalance	< 2%(1 - 100 fC)
Gain (Differential)	$\sim 5.2 \ \mathrm{mV/fC}$
Z_{IN}	110 Ω
Z_{OUT}	50 Ω
Rise time	$\sim 2 \text{ ns} (C_D = 24pF)$
Fall time	$\sim 13 \text{ ns} (C_D = 24 pF)$
Noise	1350 erms ($C_D = 24pF$)
V_{SUPPLY}	4V
P_D	$\sim 30 \mathrm{mW}$

the channel density is low and simple circuit can be used, an approach based on SMT technology can be adopted thus avoiding a specially designed (and expensive) ASIC development.

As an example, the simulation of a three



Figure 7.20: Preamplifier output for a 10, 20 and 30 fC test pulse ($C_{DET} = 24 \ pF$)

stages transimpedance preamplifier based on SiGe transistors has been carried out. The first stage dominant pole is around $26 \ MHz$ while other stages have been designed with wider bandwidth thus resulting in a good separation in terms of cutoff frequencies. Simulation results are given in table 7.3 while Fig. 7.20 shows the (simulated) output waveforms for 3 different input charges (10, 20 and 30 fC) injected through the test input.

7.4.6 Sampled Waveforms specifications

Cluster Counting technique is very powerful as it leads to an improved particles identification. The method is based on the measurement of primary ionization. To fully exploit the technique, individual electrons clusters must be identified. In our system we use slow drift gas mixtures (~ 1 $\mu s/cm$), state of the art high sampling frequency digitizers (at least 1 GSPS) and fast processing (data throughput must sustain the SuperB expected 150 kHz average trigger rate). These modules require a large amount of power forcing us to limit the number of channels to 8/16channels per Amplifier Digitizing Board (ADB). Also on the preamplifiers side fast amplifiers must be used resulting in a larger power requirements than for "Standard Readout". As a consequence, we are considering the use of a local liquid cooling system. The wide bandwidth requirement has also an impact on the type of cables used to interconnect preamplifiers and ADBs and on the full system noise pick-up sensitivity.

In addition the use of the Cluster Counting technique requires that the signal reflection in the sense wires be eliminated. This is done by means of termination resistors (R_T) that result in a lower limit on the system intrinsic noise.

Concerning the tracking requirements, if we assume 100% efficiency in the detection of each electron cluster, then the Cluster Counting dE/dx measurements includes information for tracking purposes This requires to store the arrival time of the clusters at the sense wires instead of simply counting them.

Specifications for the Sampled Waveforms measurements are the same we defined for the Standard Readout. That is: resolution, dynamic range and linearity.

7.4.6.1 Resolution

The resolution of the digitizers depends on the lowest signal amplitude to be sampled and the system baseline noise. Assuming an average input signal of ~ $6fC/e @ 3 \cdot 10^5$ gas gain, a preamplifier-shaper gain of 10mV/fC and a safety factor of 2 for gas gain fluctuations the average cluster signal is about 30 mV for a single electron.

We can estimate the preamplifier ENC from the contribution of the dominant noise source, that if of the termination resistor. Assuming a CR - RC shaping circuit and a 3 ns peaking time we get an ENC of about 0.2 fC, that is about 2 mV rms for a preamplifier gain of 10 mV/fC. Thus a voltage resolution of about 2 mV allows a good control of system noise and cluster signals reconstruction.

7.4.6.2 Dynamic range

The Cluster Counting method requires the observation of peaks (corresponding to the clusters) in the digitized signals. The signal dynamic range (discarding gas fluctuations) is then given (as an upper limit) by the expected total ionization.Helium based gas mixture have already been well characterized [7]. We can assume that a m.i.p. crossing orthogonally a 1.2 cm square cell, filled with a 90/10 He/Iso gas mixture, will generate about 22 electrons. Thus an 8 bits ADC dynamic range is fully adequate for Cluster Counting measurements.

7.4.6.3 Linearity

As we are interested in finding (and tagging) signal peaks, a resolution of 2% fully satisfies the requirements.

7.4.7 Sampled Waveforms - DCH Front End Electronics (overall design)

The Sampled Waveforms DCH FEE chain is similar to the Standard Readout one. In this scenario we will have electronics modules connected by means of mini coaxial cables. Because of the smaller number of channels per ADB both the number of crates and ADBs will increases significantly (tables 13.4 and 13.6).

7.4.8 Sampled Waveforms - On Detector Electronics

7.4.8.1 Preamplifier Boards

Because preamplifier boards will host high bandwidth (~ 350 MHz) amplifiers the layout and assembly are more difficult compared to the Standard Readout scenario. Particularly, special attention must be provided to avoid ground loops to minimize instabilities and external noise pickup.

Something about preamplifier

7.5 High Voltage system Robertson, Martin 1 page

- 7.5.1 Main HV system and cable routing - Robertson 0.5 page
- 7.5.2 Distribution boards Martin 0.5 page
- 7.5.3 HV distribution boards Standard ReadOut

The high voltage distribution network will be located on the forward end-plate. The distribution board modularity will match the preamplifier modularity while the number of distribution boards connected to a single HV channel will depend on the layer (example: inner layers = 2 boards, outer layers = 5 boards).

The HV distribution system consists of power supplies and distribution boards, along with associated cables, feedthroughs into the forward electronics enclosure volume and HV distribution within that volume. The voltage will be supplied by a CAEN SY4527 Universal Multichannel System supply with 16 A1535N distribution boards, giving a total of 384 channels. This is sufficient granularity that a single channel failure will have a small impact on detector performance. Spares of both the SY4527 and the A1535N will be on hand. The A1535N permits individual channels to operate in currentgenerator mode in case of over-currents. This feature was found to be extremely useful for the BABAR drift chamber as it permitted the chamber to handle locally high background rates without ramping down the chamber HV.

Individual HV channels are brought to the drift chamber from the A1535N boards via multistrand cables terminating at both ends with A996 52 pin Radiall connectors.

The multiconductor cable connects to a filter box containing a low-pass filter, located at the inner radius of the forward enclosure. The individual channels are fanned out within the enclosure to the HV distribution boards (Sec. 7.3.6). Each HV channel supplies two or three 8-channel distribution boards, depending on the superlayer.

The HV distribution and sense wire termination circuitry is shown in Fig. 7.21. If termination is not used, the termination resistor (R_T) and the 500 pF capacitor per sense wire are not needed.



Figure 7.21: HV distribution network.

7.6 Gas system

142

The drift chamber is filled with a gas mixture of 90% helium and 10% isobutane. The gas system supplies the appropriate gas mixture to the chamber, while maintaining the required flow rate, pressure, purity, and composition stability. It includes safety items such as flammable gas sensors and release valves to protect the gas system, detector, and personnel from dangerous conditions caused by component failures or operator errors. To reduce the operating costs, 85% of the gas will be purified and recirculated.

The mixing will be done using mass flow controllers that will maintain the isobutane fraction at $(10.0 \pm 0.1)\%$. A parallel set of rotameters may be included to allow for high flow rate flushing. The system will allow a fraction of the flow to pass through a temperature-controlled water bath, which will be used if we decided to add water to help control the symptoms of aging.

The gas composition is verified using a set of analyzers to measure isobutane, oxygen, and water. The analyzer set will be able to sample gas at a variety of points in the gas system, such as before or after the gas enters the chamber or the filters.

This will be a recirculating gas system, which reduces operating costs and air pollution from the isobutane. The total flow will be 15 liters per minute, of which 2.5 liters per minute will be fresh gas. For a chamber volume of 5000 liters, this corresponds to four volume changes per day, or one volume of fresh gas every 1.5 days. The flow is controlled by an explosionproof compressor, which is regulated to maintain a chamber pressure of 4.00 ± 0.05 mbar (0.4% of an atmosphere) above atmospheric pressure.

The primary gas lines between the gas mixing station and the detector will be welded and pressure-tested stainless pipe, 1.5 inches diameter. This will be reduced to 0.75 inch diameter in the cable trays through the detector. The input line is fanned out to 8 lines of 5 mm diameter on the rear endplate, while the output line is fanned out to 16 lines of 5 mm diameter on the forward endplate.

The gas returning from the detector passes through a palladium catalytic filter which removes oxygen by the reaction $13O_2 + 2C_4H_{10} \rightarrow 8CO_2 + 10H_2O$. The resulting water is removed by an alumino-silicate molecular sieve. The system contains two such sieves, so that one can be regenerated (i.e., have the absorbed water removed) by flushing with helium at elevated temperature without stopping operations. This filter system was originally built for the *BABAR* drift chamber and will be reused for SuperB.

The gas temperature and pressure will be monitored at various points in the system, along with atmospheric pressure. These quantities will be used to calculate gain correction due to gas density. We will also monitor gas gain using a small, single-cell chamber that will be mounted on the return line from the chamber. Any variations in the current induced by an 55 Fe source after applying the density correction would indicate gain variations due to gas composition or chamber aging.

The majority of the gas system components will be in the gas hut (or room), which will be located at an exterior wall of the interaction hall. Two additional racks close to the detector will contain bubblers, pressure sensors, and valves. The gas storage areas will be outside, under cover, immediately adjacent to the hut. The isobutane, since it is flammable, will be stored in a physically separate area from the other gases. The isobutane tanks and lines will be heated and insulated.

The gas system includes an extensive safety system to protect personnel and equipment. This system will be reviewed and approved by the laboratory. Aspects of the safety system include:

- ventilation in the gas hut, which, when combined with flow restrictors on the lines into the hut, ensure that a leak cannot create an asphyxiation hazard.
- nitrogen flows in the exhaust lines and into the electronics enclosures on both endplates.
- flammable gas sensors in the gas hut and the bubbler rack.

- an oxygen sensor on the return line.
- bubblers and redundant pressure sensors to protect the chamber against over pressure.
- an independent helium line and regulator to protect the chamber against subatmospheric pressures.
- administrative controls on changes to the gas system.

The system is designed such that it will remain safe even during extended power outages. We will undertake regular maintenance and keep sufficient spare parts to ensure reliable operations.

7.7 Calibration and monitoring -Roney 3 pages

- 7.7.0.1 Slow control systems
- 7.7.0.2 Calibration
- 7.7.0.3 Gas monitoring system
- 7.7.0.4 On-line monitor
- 7.8 Integration

7.8.1 Overall geometry and mechanical support

The envelope of the drift chamber is determined by the tungsten shield and the DIRC at the inner and outer radii, and by the backward calorimeter and the FTOF in the negative and positive z directions. There are 5 mm radial clearances between the drift chamber and the surrounding components, and 5 mm clearance between the drift chamber envelope and the backward calorimeter. The FTOF is directly mounted onto the drift chamber. The envelope in the backward direction includes the space occupied by the signal cables after they exit the enclosure.

In BABAR, the chamber was supported at the backward end by turnbuckles connecting the

rear cylinder extension to the DIRC strong support tube. We envision using a similar system, although the actual mounting points used by *BABAR* will be obscured by the backward calorimeter.

In the forward direction, the drift chamber and FTOF form an integrated mechanical package, supported by the DIRC central support tube (CST). Figure. 7.22 shows the forwards mounting components used for the *BABAR* drift chamber. Note that the support point on the DIRC is on the z surface of the CST, not the inner radius. Because the Super*B* chamber is shorter in the forward direction than *BABAR*, the corresponding support tabs will be on the FTOF, not the chamber.

7.8.2 Installation and alignment

The chamber and FTOF will be installed prior to the forward and backward calorimeters and the tungsten shielding. The installation will reuse the existing *BABAR* equipment, which is currently stored at SLAC. The chamber is supported at the inner radius and slid along a supporting beam that passes through the inner cylinder (Fig. 7.23).

Both forward and backward enclosures will contain a number of precision 6 mm dowel into which target holders for corner-cube reflectors can be mounted. The enclosures in turn are doweled to precise reference holes on the endplates, referencing the target locations to the sense wire locations.

The mounting systems at both ends allow for several mm of adjustment in x and y. The chamber location will be adjusted to center the chamber in x and y on the interaction point and to align the sense wire direction with the magnetic field. The tolerances on these alignments have not yet been specified. The tolerance on the location in z will be significantly looser.

7.8.3 Services

The services required for the backward end are listed below. These will reach the backward enclosure via 16 slots in the outer radius of the steel plug located within the DIRC strong tube. Each slot will be approximately 50 mm in radius



Figure 7.22: Forward end mounting system on the BABAR drift chamber. (a) Tab on the outer radius of the drift chamber. (b) Corresponding mounting point on the CST.

by 250 mm wide. Cables continuing to the digitizing crates, located on the top of the detector, will be routed to wireways at the end of the IFR iron immediately after exiting the DIRC strong tube.

Note that within the radial extent of the backward calorimeter, the signal cables (and all other services) are routed to stay within the drift chamber envelope.

• Signal cables: approximately 8000 coaxial cables, RG-179, 2.54 mm in diameter, organized into 8-cable ribbons. These are 10 m in length, and travel to the digitizing electronics crates.



Figure 7.23: BABAR drift chamber during installation. The same tooling is available for use by SuperB.

- Calibration cables: RG-179, one for every eight signal channels. Also originate at the digitizer crates.
- Low-voltage power: Four 1/0 welding cables, 14.7 mm diameter. Originate in the electronics hut.
- Cooling lines: 16 lines (8 separate circuits, each with a supply and a return line), 26.2 mm reinforced PVC. The subatmospheric water-based cooling system (Sec. 7.3.6) will be close to the detector, but outside of the radiation area.
- Drift gas: one line, 19 mm diameter stainless steel. Originates in the gas hut. This line increases to 38 mm diameter after exiting the detector.
- Nitrogen flush gas: two lines, 19 mm diameter stainless steel. Also from the gas hut.

The services for the forward end will exit the detector in the radial space between the tungsten shield and the FCAL. The services for the forward region are:

16 56-conductor cables, • High voltage: 14 mm diameter. Originate at the HV supplies in the electronics hut.

145

- Drift gas: one line, 19 mm diameter stainless steel. Originates in the gas hut. This line increases to 38 mm diameter after exiting the detector.
- Nitrogen flush gas: two lines, 19 mm diameter stainless steel. From the gas hut.
- Cooling lines: 2 lines, 19 mm reinforced PVC. From the cooling system.

7.9 Bibliography

- R. Veenhof, Garfield simulation of gaseous detectors CERN-W5050, Version 9.
- [2] CLEO reference.
- [3] M. Adinolfi et al. (KLOE Collaboration), The tracking detector of the KLOE experiment, Nucl. Instrum. Methods Phys. Res., Sect. A 488, 51 (2002).
- [4] M. Adinolfi et al. (KLOE Collaboration), The tracking detector of the KLOE experiment, Nucl. Instrum. Methods Phys. Res., Sect. A 488, 51 (2002).
- [5] A. Sharma, F. Sauli, Nucl. Instr. and Meth. A 350 (1994) 470.
- [6] C. Avanzini et Al. Nuclear Instruments and Methods in Physics Research A449 (2000). 237-247.
- P.R.Burchat and John Hiser Studies of helium gas mixtures in drift chambers SLAC-PUB-5626 SCIPP 91/25 September 1991.