



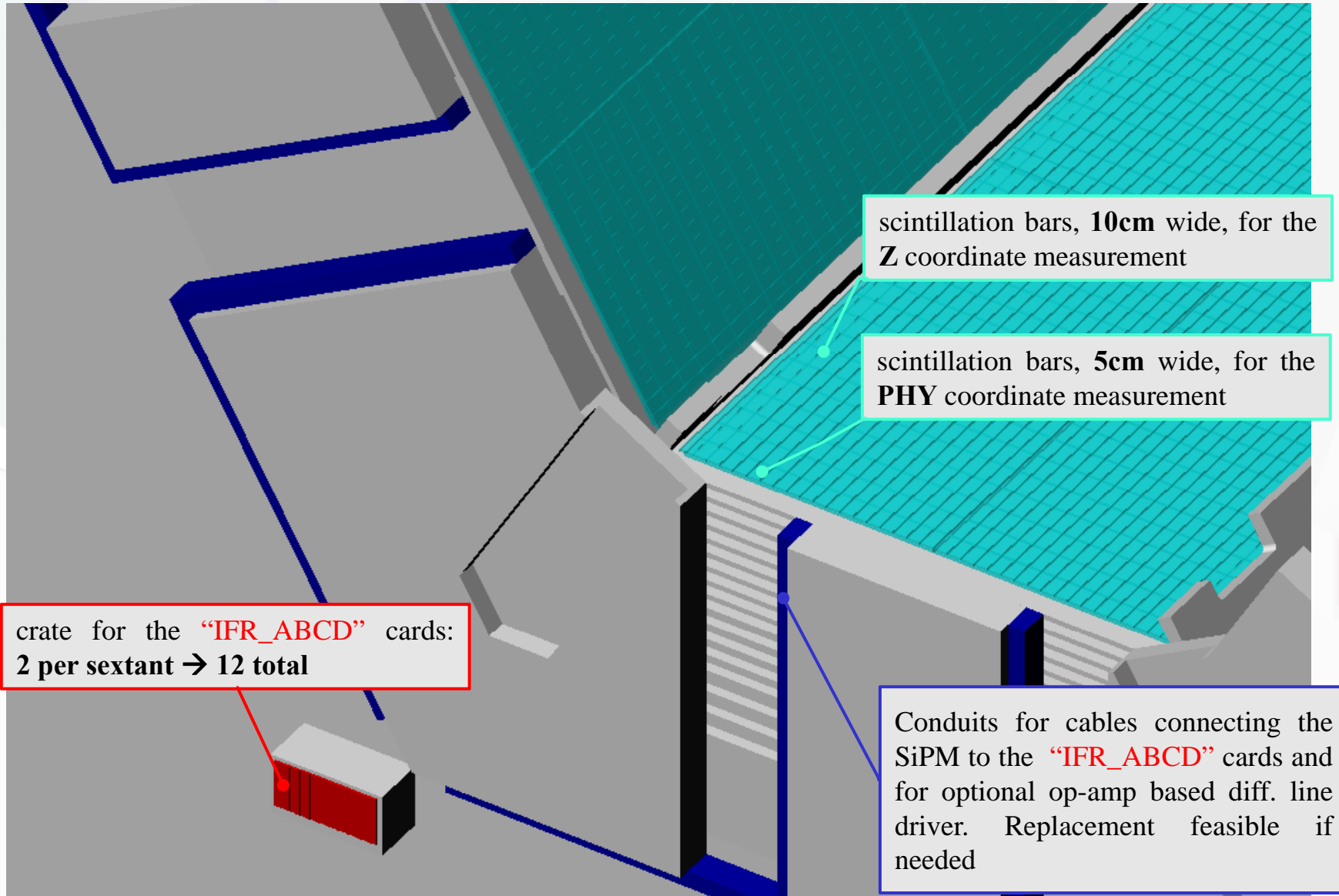
SuperB IFR electronics: report at the SuperB ETD meeting

a.c.f. 2011-11-02

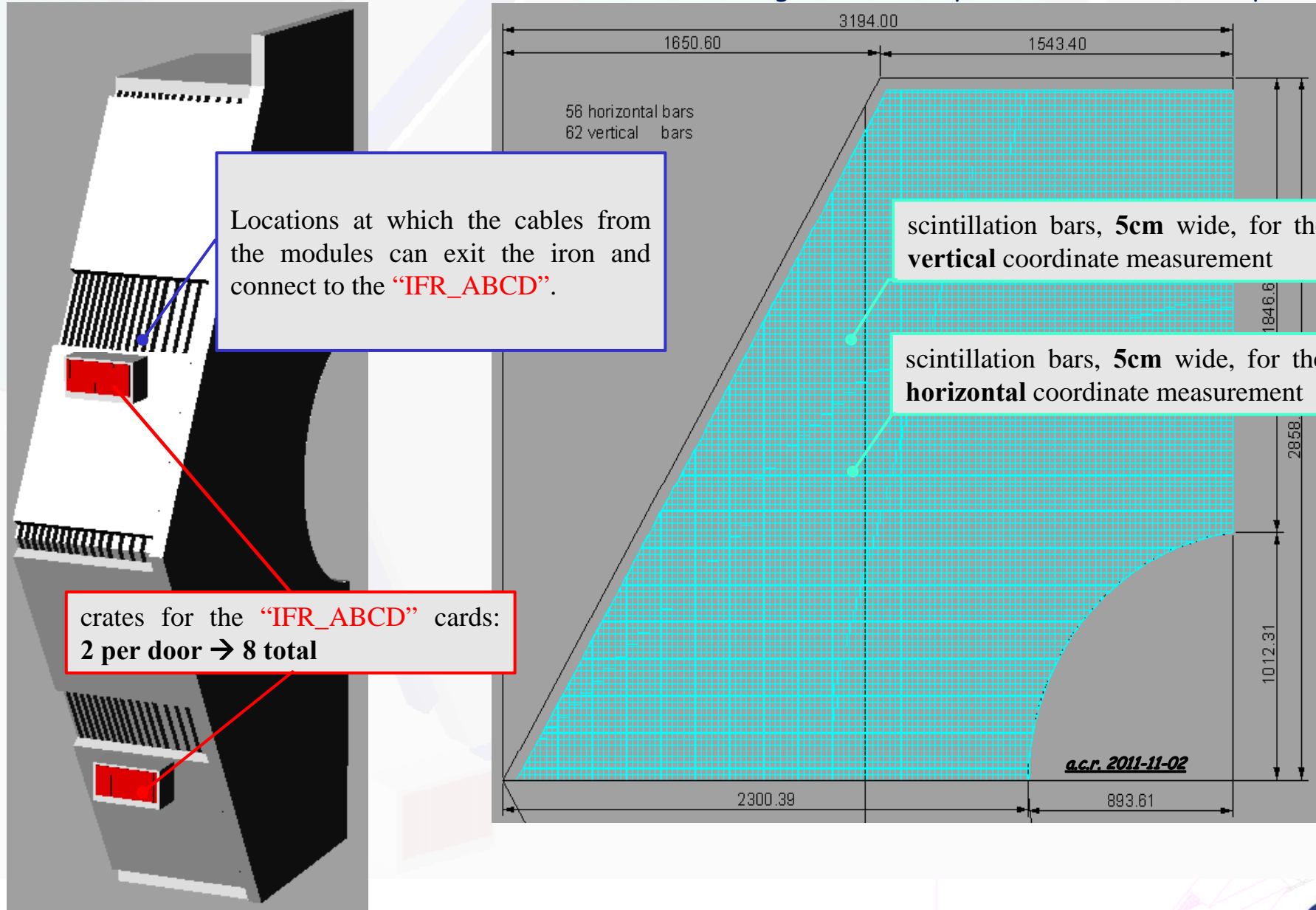
## SuperB IFR electronics: report at the SuperB ETD meeting

- features of the current baseline IFR detector design with “binary mode” readout
  - layout of the detector elements in the barrel
  - layout of the detector elements in the endcaps
  - passive Single Ended vs active Differential options for “picking-up” the signals from SiPM
  - block diagram of the “**IFR\_ABCD**” front end card
- channel number estimation
  - barrel
  - endcaps
- data rate / data link count estimation
  - at nominal conditions (trigger rate 150KHz)
  - at higher luminosity (trigger rate 500KHz)
- addressing the radiation issues
  - background rates estimates at the location of IFR front end electronics
  - measurement of neutron induced SEE in FPGA at the INFN Laboratori Nazionali di Legnaro (exploiting beam time scheduled for SiPM test; thanks to Roberto Stroili, Flavio Dal Corso, INFN Padova)
  - measurement of TID and SEE on the optional front end op-amps and the EASIROC at the INFN Laboratori Nazionali di Legnaro (being planned)

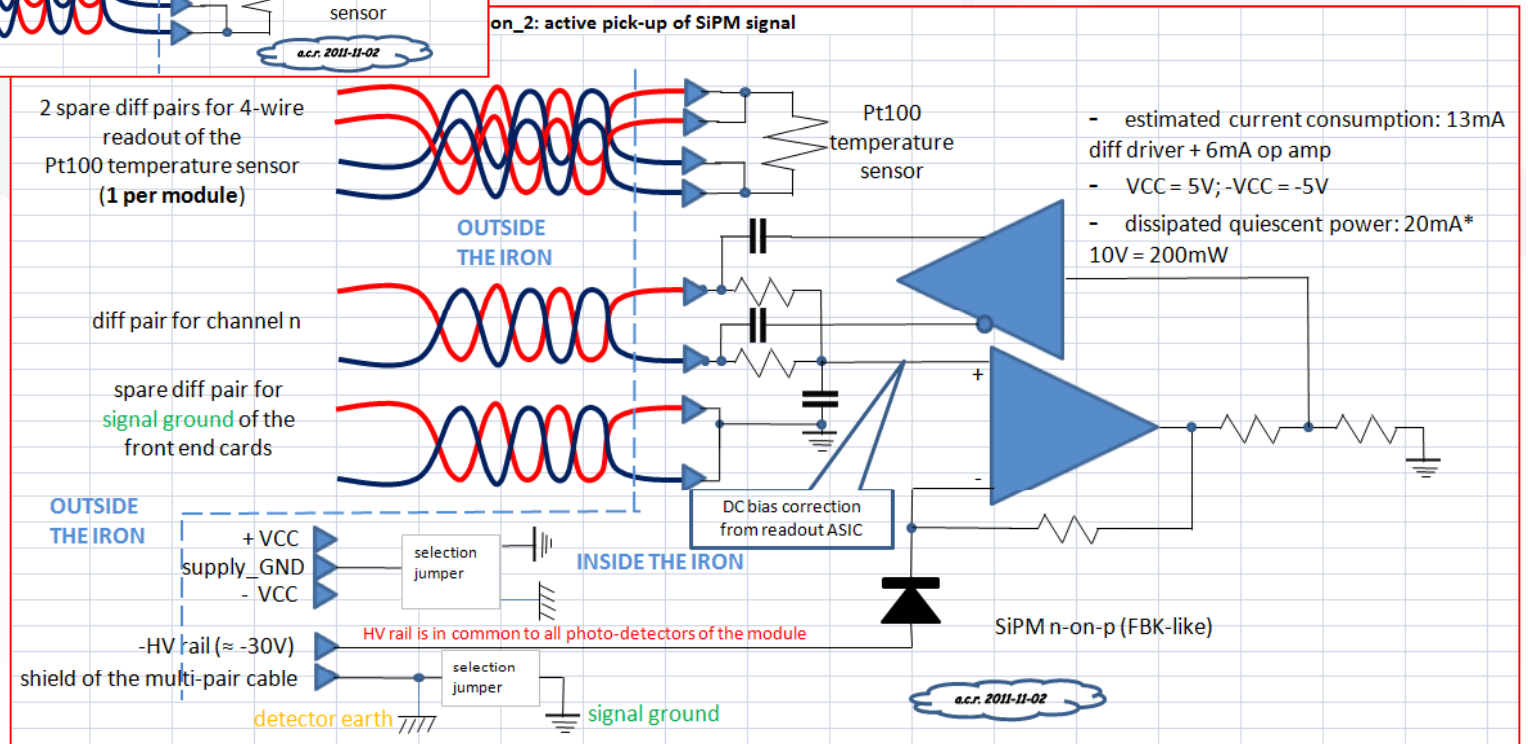
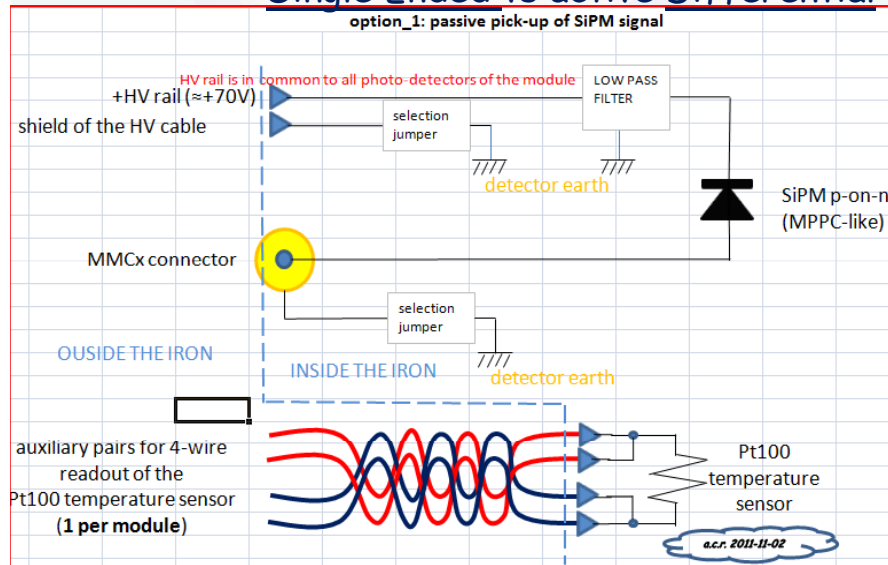
features of the current baseline IFR detector design with "binary mode" readout: barrel



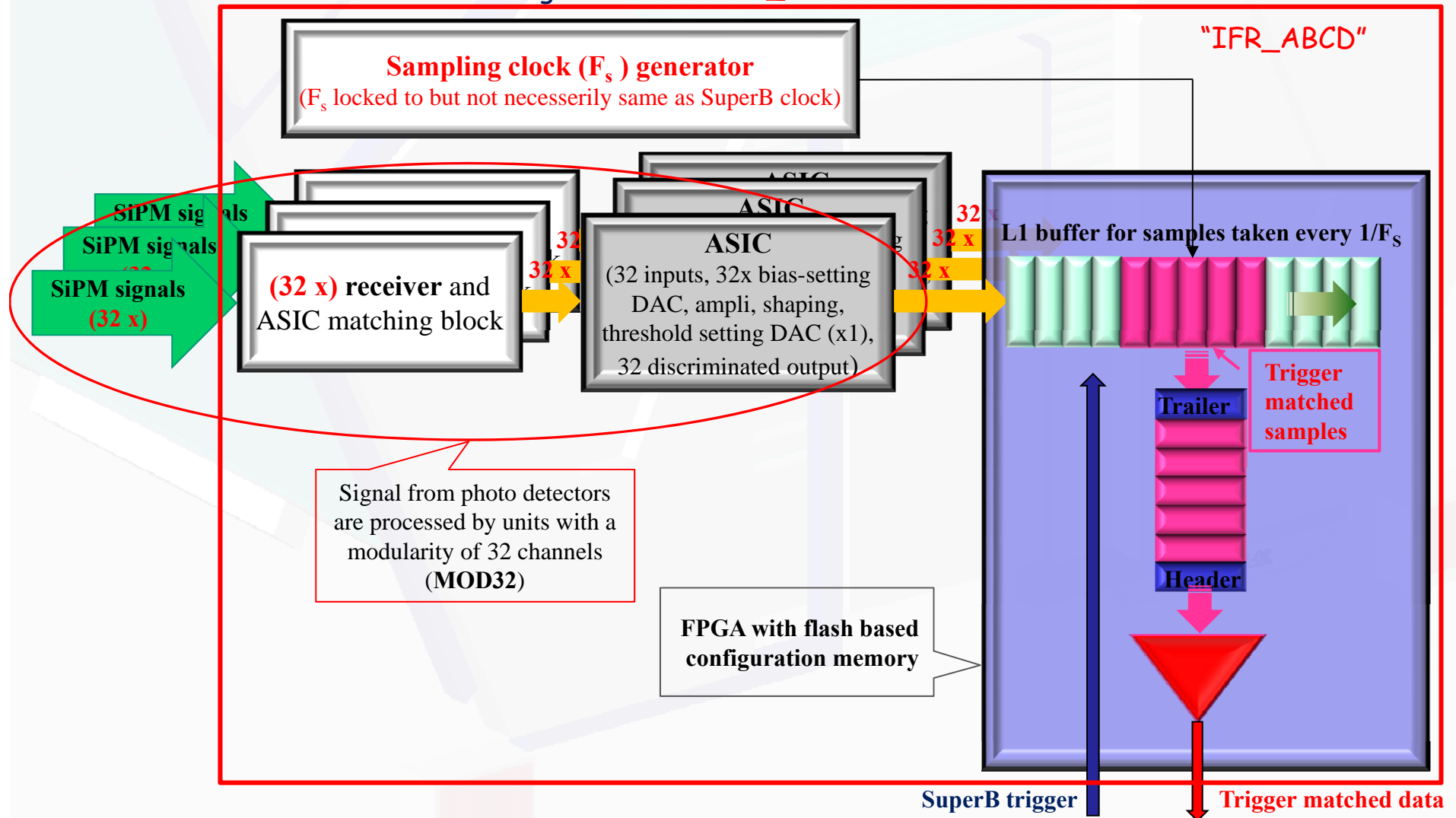
# features of the current baseline IFR detector design with "binary mode" readout: endcaps



# features of the current baseline IFR detector design with "binary mode" readout: Single Ended vs active Differential options for "picking-up" the signals from SiPM

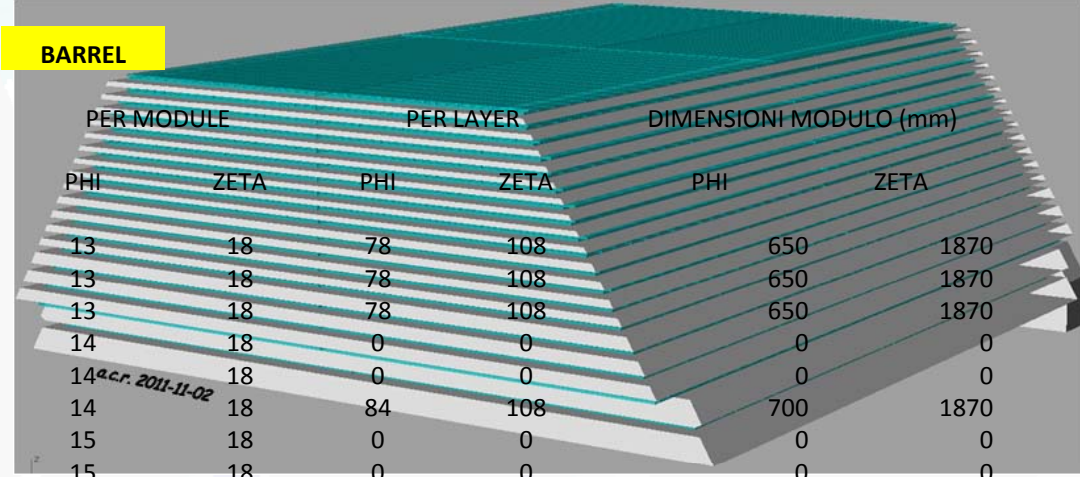


features of the current baseline IFR detector design with "binary mode" readout:  
 block diagram of the "IFR\_ABCD" front end card



Two "CONCENTRATOR / DATA LINK DRIVER" cards in the "IFR-ABCD" front end crates are connected to the FCTS and to the ROM: they distribute the trigger to the "IFR\_ABCD" cards, collect the matched data, merge the streams into the de-randomizer FIFO and drive the data link (1 or 2 per concentrator) toward the ROMs

# channel number estimation: barrel



LAYER WIDTH	LAYER	No.Modules per layer	LAYER ENABLE
1963	1	6	1
1987	2	6	1
2050	3	6	1
2113	4	6	
2176	5	6	
2240	6	6	1
2304	7	6	
2367	8	6	
2431	9	6	
2494	10	8	1
2569	11	8	
2641	12	8	
2712	13	8	1
2784	14	8	
2879	15	8	1
2973	16	8	
3068	17	8	
3144	18	8	1
3296	19	8	1

PER MODULE		PER LAYER		DIMENSIONI MODULO (mm)	
PHI	ZETA	PHI	ZETA	PHI	ZETA
13	18	78	108	650	1870
13	18	78	108	650	1870
13	18	78	108	650	1870
14	18	0	0	0	0
14 <sup>a.c.r. 2011-11-02</sup>	18	0	0	0	0
14	18	84	108	700	1870
15	18	0	0	0	0
15	18	0	0	0	0
16	18	0	0	0	0
12	18	96	144	600	1870
12	18	0	0	0	0
13	18	0	0	0	0
13	18	104	144	650	1870
13	18	0	0	0	0
14	18	112	144	700	1870
14	18	0	0	0	0
15	18	0	0	0	0
15	16	120	128	750	1870
16	16	128	128	800	1870

NUMBER OF MODULES per sextant: 134

**TOTAL NUMBER OF MODULES: 804**

TOTAL PER SEXTANT: 9

TOTAL PER SEXTANT: 1998, 878, 1120

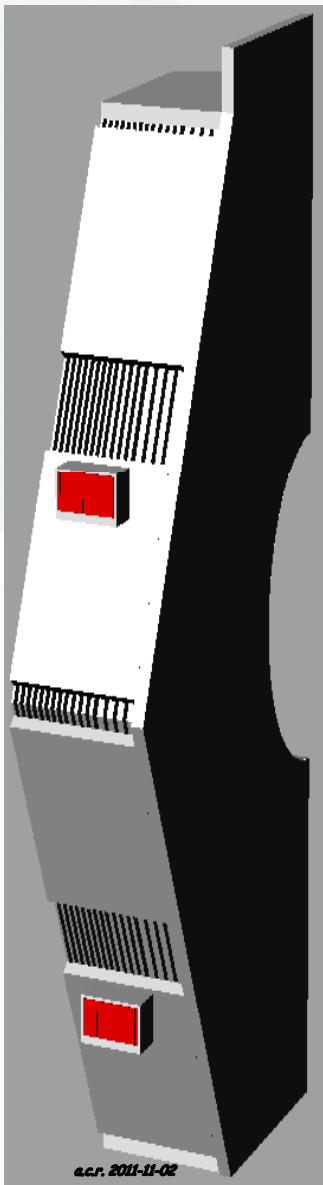
NUMBER OF SEXTANTS: 6

TOTAL PER BARREL: 54

**TOTAL CHANNELS PER BARREL: 11988**



## channel number estimation: endcaps



horizontal bars per module:	56
vertical bars per module:	62
channel count per module:	118
NO_OF_MOD_PER_LAYER_EC:	2
horizontal bars per layer	112
vertical bars per layer	124
NUMBER OF LAYER PER DOOR	9
NUMBER OF DOORS IN ENDCAPS	4

### ENDCAP

**TOTAL NUMBER OF MODULES IN ENDCAPS: 72**

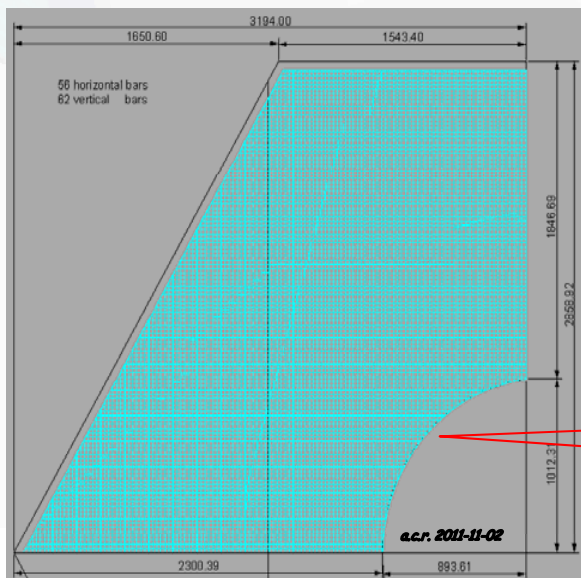
horizontal bars per door: 1008

vertical bars per door: 1116

TOTAL HORIZONTAL BARS: 4032

TOTAL VERTICAL BARS: 4464

**TOTAL CHANNELS IN ENDCAPS: 8496**



Dimensions of one endcap module



## data rate / data link count estimation: at nominal conditions (trigger rate 150KHz)

BARREL				ENDCAP			
NUMBER OF CRATES PER SEXTANT	2	NUMBER OF CHANNELS PER CRATE	999	NUMBER OF CRATES PER DOOR	2	NUMBER OF CHANNELS PER CRATE	1062
NUMBER OF BOARDS PER CRATE	12	NUMBER OF CHANNEL PER BOARD	83,25	NUMBER OF BOARDS PER CRATE	12	NUMBER OF CHANNELS PER BOARD	88,5
NUMBER OF DATA LINKS PER CRATE (equal to number of data concentrator/derandomizer buffers)	2	NUMBER OF 32-CHANNEL PROCESSING UNITS	3	NUMBER OF DATA LINKS PER CRATE (equal to number of data concentrator/derandomizer buffers)	2	NUMBER OF 32-CHANNEL PROCESSING UNITS (MOD32) PER BOARD	3
TRIGGER RATE (kHz)		150		TRIGGER RATE (kHz)		150	
SAMPLING FREQUENCY (MHz)	50	SAMPLING FREQUENCY (MHz)	50	SAMPLING FREQUENCY (MHz)	50	SAMPLING FREQUENCY (MHz)	50
SAMPLING CLOCK PERIOD (ns)	20	SAMPLING CLOCK PERIOD (ns)	20	SAMPLING CLOCK PERIOD (ns)	20	SAMPLING CLOCK PERIOD (ns)	20
W: READOUT WINDOW (ns)		150		W: READOUT WINDOW (ns)		150	
n: NUMBER OF SAMPLES PER TRIGGER		8		n: NUMBER OF SAMPLES PER TRIGGER		8	
NUMBER OF HEADER/TRAILER WORDS	2	NUMBER OF HEADER/TRAILER WORDS	2	NUMBER OF HEADER/TRAILER WORDS	2	NUMBER OF HEADER/TRAILER WORDS	2
No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS (MOD32)	10	No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS (MOD32)	10	No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS (MOD32)	10	No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS (MOD32)	10
EVENT SIZE IN BYTES PER EACH MOD32 UNIT	40	TOTAL NUMBER OF MOD32 UNITS	432	EVENT SIZE IN BYTES PER EACH MOD32 UNIT	40	TOTAL NUMBER OF MOD32 UNITS	288
BARREL EVENT SIZE in KB (TOTAL NUMBER OF BYTES PER EVENT FOR ALL MOD32 UNITS)		17280		ENDCAP EVENT SIZE in KB (TOTAL NUMBER OF BYTES PER EVENT FOR ALL MOD32 UNITS)		11520	
TOTAL BARREL BANDWIDTH (MB/s)	2592	TOTAL BARREL BANDWIDTH (Gbps) assuming 8b/10b encoding overhead	25,92	TOTAL ENDCAP BANDWIDTH (MB/s)	1728	TOTAL ENDCAP BANDWIDTH (Gbps) assuming 8b/10b encoding overhead	17,28
TOTAL NUMBER OF DATA LINK		24		TOTAL NUMBER OF DATA LINK		16	
		bandwidth per link (Gbps)				bandwidth per link (Gbps)	
		1,08				1,08	

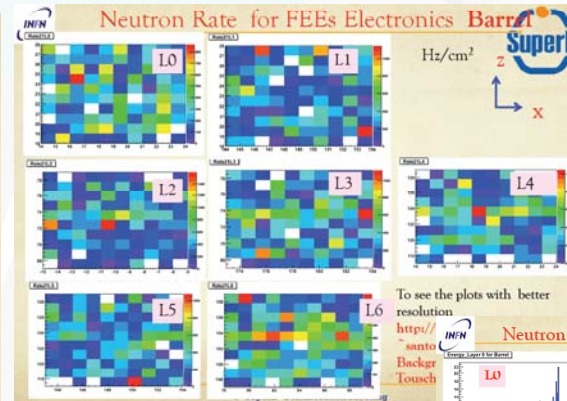
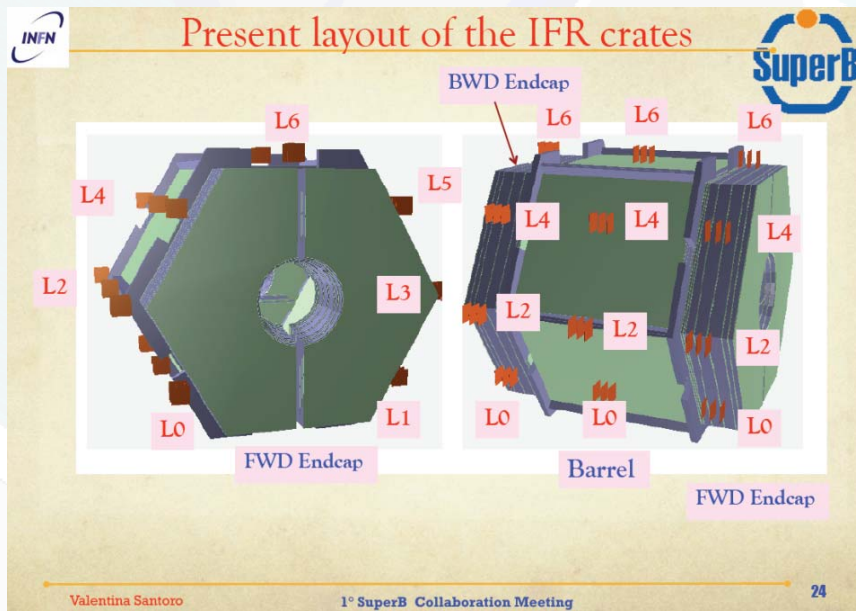
## data rate / data link count estimation: at higher luminosity (trigger rate 500kHz)

BARREL			
NUMBER OF CRATES PER SEXTANT	2	NUMBER OF CHANNELS PER CRATE	999
NUMBER OF BOARDS PER CRATE	12	NUMBER OF CHANNEL PER BOARD	83,25
NUMBER OF DATA LINKS PER CRATE (equal to number of data concentrator/derandomizer buffers)	4	NUMBER OF 32-CHANNEL PROCESSING UNITS	3
<b>TRIGGER RATE (kHz) 500</b>			
SAMPLING FREQUENCY (MHz)	50		
SAMPLING CLOCK PERIOD (ns)	20		
<b>W: READOUT WINDOW (ns)</b>	<b>150</b>		
<b>n: NUMBER OF SAMPLES PER TRIGGER</b>	<b>8</b>		
NUMBER OF HEADER/TRAILER WORDS	2		
No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS ( <b>MOD32</b> )	10		
EVENT SIZE IN BYTES PER EACH <b>MOD32</b> UNIT	40	TOTAL NUMBER OF <b>MOD32</b> UNITS	432
<b>BARREL EVENT SIZE in KB</b> (TOTAL NUMBER OF BYTES PER EVENT FOR ALL <b>MOD32</b> UNITS)		<b>17280</b>	
TOTAL BARREL BANDWIDTH (MB/s)	8640	TOTAL BARREL BANDWIDTH (Gbps) assuming 8b/10b encoding overhead	86,4
<b>TOTAL NUMBER OF DATA LINK 48</b>		bandwidth per link (Gbps)	<b>1,8</b>

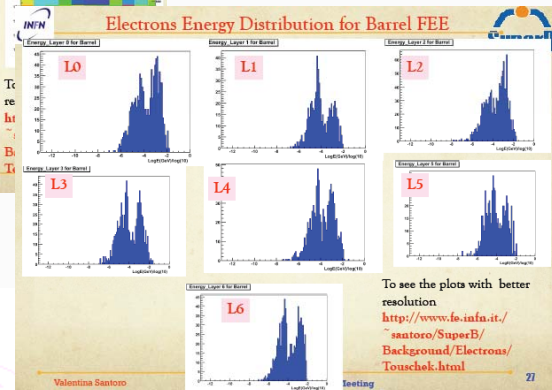
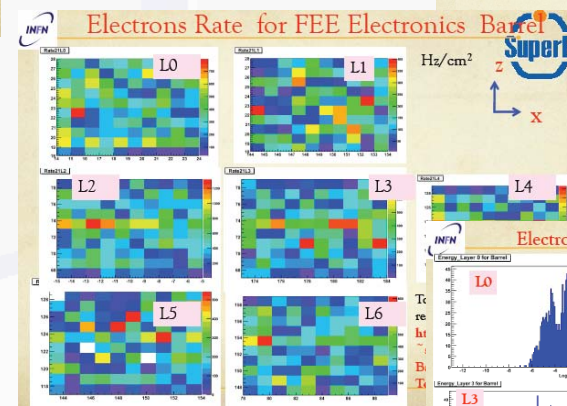
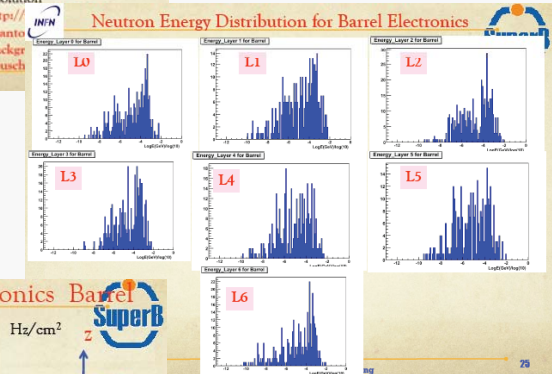
ENDCAP			
NUMBER OF CRATES PER DOOR	2	NUMBER OF CHANNELS PER CRATE	1062
NUMBER OF BOARDS PER CRATE	12	NUMBER OF CHANNELS PER BOARD	88,5
NUMBER OF DATA LINKS PER CRATE (equal to number of data concentrator/derandomizer buffers)	4	NUMBER OF 32-CHANNEL PROCESSING UNITS ( <b>MOD32</b> ) PER BOARD	3
<b>TRIGGER RATE (kHz) 500</b>			
SAMPLING FREQUENCY (MHz)	50		
SAMPLING CLOCK PERIOD (ns)	20		
<b>W: READOUT WINDOW (ns)</b>	<b>150</b>		
<b>n: NUMBER OF SAMPLES PER TRIGGER</b>	<b>8</b>		
NUMBER OF HEADER/TRAILER WORDS	2		
No. OF 32 BIT WORDS PER TRIGGER PER 32-CHANNEL PROCESSING UNITS ( <b>MOD32</b> )	10		
EVENT SIZE IN BYTES PER EACH <b>MOD32</b> UNIT	40	TOTAL NUMBER OF <b>MOD32</b> UNITS	288
<b>ENDCAP EVENT SIZE in KB</b> (TOTAL NUMBER OF BYTES PER EVENT FOR ALL <b>MOD32</b> UNITS)		<b>11520</b>	
TOTAL ENDCAP BANDWIDTH (MB/s)	5760	TOTAL ENDCAP BANDWIDTH (Gbps) assuming 8b/10b encoding overhead	57,6
<b>TOTAL NUMBER OF DATA LINK 32</b>		bandwidth per link (Gbps)	<b>1,8</b>

addressing the radiation issues:  
background rates estimation at the location of IFR front end electronics

From: "IFR Background Status" Valentina Santoro INFN Ferrara



To see the plots with better resolution  
<https://www.infn.it/~santoro/SuperB/Background/Electrons/Touschek>

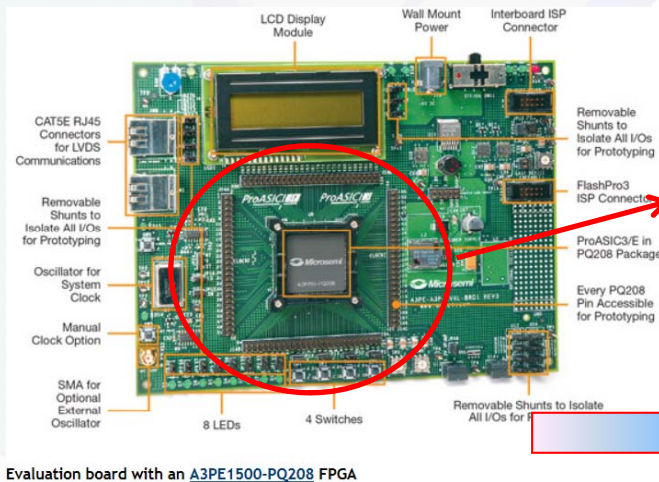


To see the plots with better resolution  
<http://www.infn.it/~santoro/SuperB/Background/Electrons/Touschek.html>

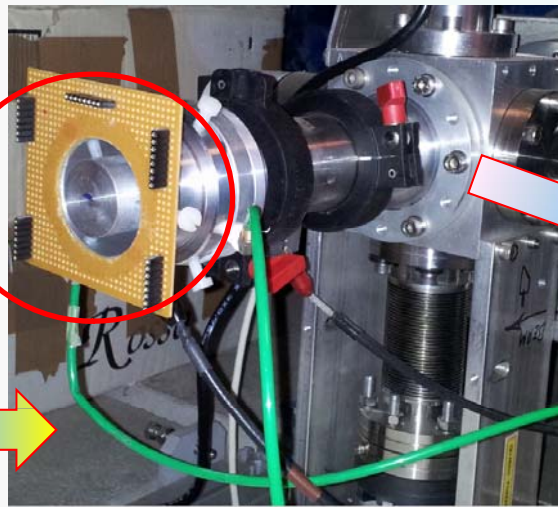
Thanks to this work we have now an estimate of the expected radiation doses at the locations of the front end electronics crates, according to the baseline design.

This will allow us to properly plan the irradiation tests of the "compact" front end design based on the EASIROC.

addressing the radiation issues:  
 measurement of neutron induced SEE in FPGA at the INFN Laboratori Nazionali di Legnaro

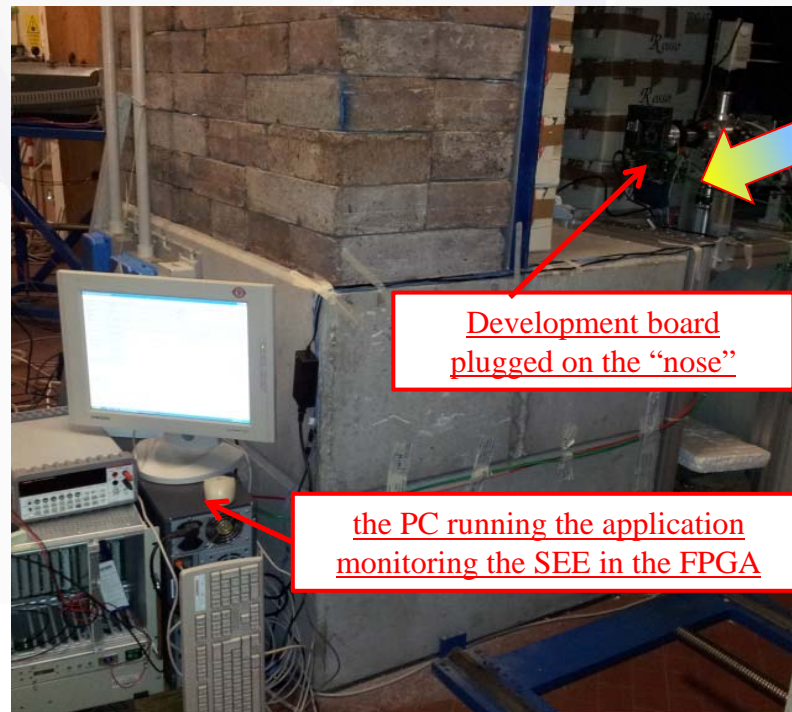


Evaluation board with an [A3PE1500-PQ208](#) FPGA



Development board plugged on the "nose"

Development board plugged on the "nose"

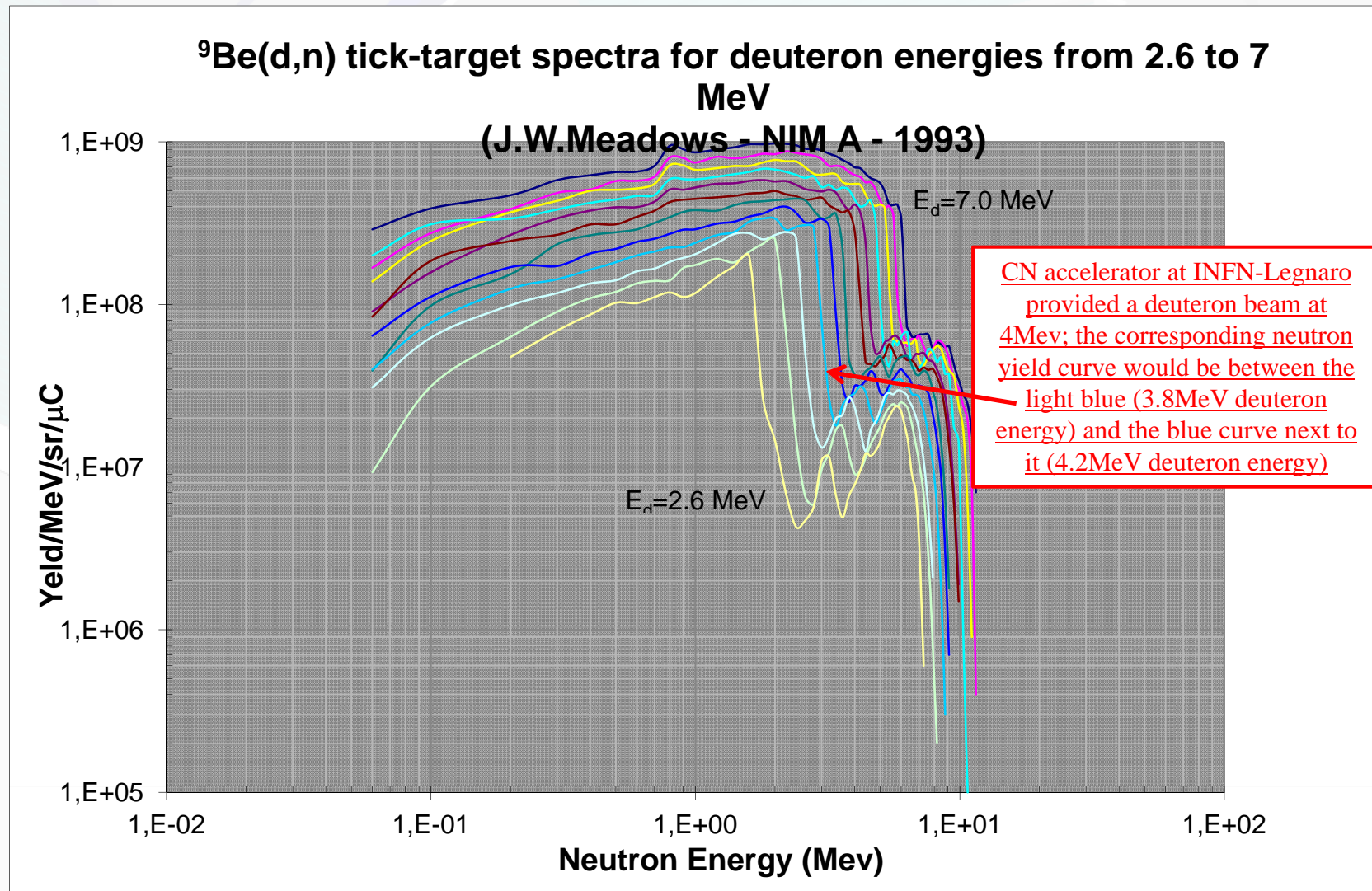


the PC running the application monitoring the SEE in the FPGA

**Acknowledgements:**

We exploited beam time which had been allocated by our colleagues Roberto Stroili and Flavio Dal Corso who also provided the information needed to estimate the neutron flux.

addressing the radiation issues:  
measurement of neutron induced SEE in FPGA at the INFN Laboratori Nazionali di Legnaro



( excel file by Flavio Dal Corso )

## addressing the radiation issues:

measurement of neutron induced SEE in FPGA at the INFN Laboratori Nazionali di Legnaro

Acknowledgements:

The test design developed as his undergraduate thesis work by **Lorenzo De Santis** (University of Ferrara) in the ACTEL device occupied the FPGA for about 98%.

It implemented:

- an embedded FIFO 4096 x 8 based on SRAM cellsm, for a total of **32768 SRAM bits**
- a shift register with 728 cells 8bit wide, for a total of **5824 Flip Flops**
- a UART to load and read back the memory blocks

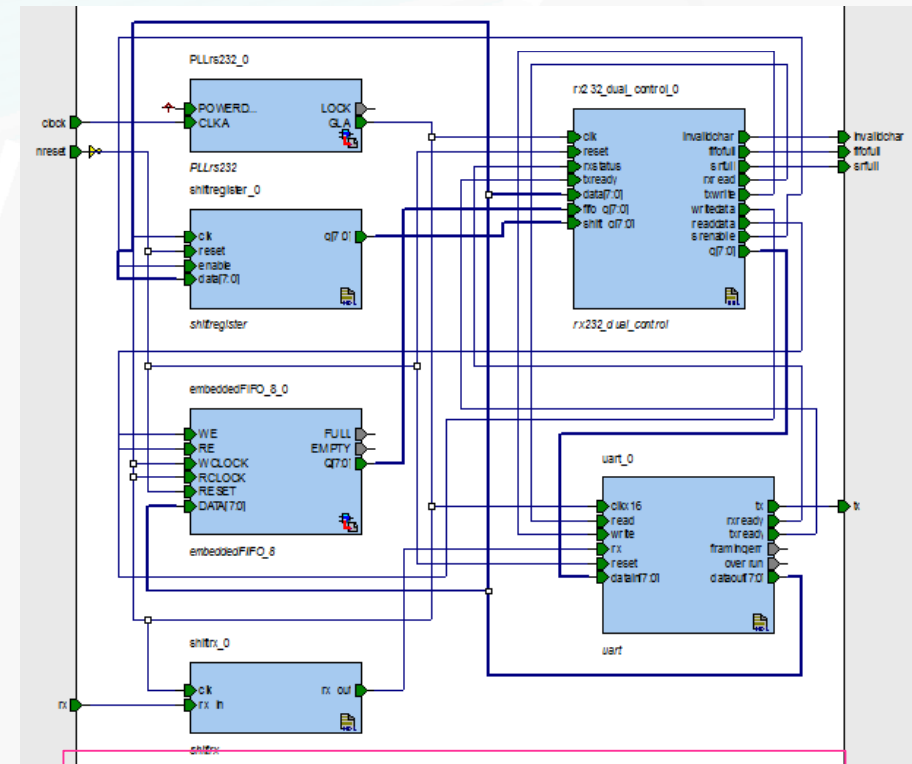
A PC running a Labview application also provided by Lorenzo was used to repeatedly run the routine of loading known patterns into the structures and reading them back after the test interval.

A typical outcome of the test was that in  $\frac{1}{2}$  hour run with 4MeV deuteron energy and 40nA average current we would detect:

- around 36 SRAM bit flip
- between 0 and 2 FLIP FLOP upsets

If our estimation of neutron flux at the test facility are correct then for each  $\frac{1}{2}$  hour run at 40nA the FPGA has been irradiated with  $\approx$  order of  $1 \cdot 10^{10}$  neutrons  $\rightarrow$

- we could use a flash based FPGA to handle L1 buffering
- we should protect the state machines with TRM but not necessarily the L1 buffer itself: data corruption rate is low



**NO bit flips were detected (nor expected) in the configuration memory (flash-based) of the FPGA: the functionality of the chip was never found impaired**

# addressing the radiation issues: measurement of neutron induced SEE in FPGA at the INFN Laboratori Nazionali di Legnaro

FIFO depth: 4096 number of bits: 32768  
 Shift Reg. Depth: 728 number of FF: 5824  
 word width: 8  
 fluenza teorica at LNL per deutoni 4MeV a.c.r.: provo a usare 1e8 che mi sembra piu' appropriato dato lo spettro **1,02E+08 neutroni/sterad \* uC**  
 (Flavio Dal Corso):

IPOTESI: h - distanza tra sorgente e silicio  
 FPGA (mm): 7,00E+00mm  
 IPOTESI: a - spigolo del die della  
 FPGA (mm): 1,00E+01mm  
 angolo solido del Si dalla sorgente (Lorenzo):  
 $4 * \arcsin[ a^2 / (a^2 + 4 * h^2) ]$  1,38E+00sterad

corrente di fascio (nA) 4,00E+01nA  
 tempo di misura (s) 1,80E+03s

numero di neutroni attraverso FPGA in 1/2 ora. E' uguale alla fluenza se assumo che area FPGA = 1 cm^2 **1,01E+10 neutrons thru FPGA in 1/2 hour**

<numero di bit flip in FIFO> : 37,0000  
 per Mb: 1129,1504  
 <numero di bit flip in Shift Reg> : 1,5000  
 per Million Flip Flop: 257,5549

vedi: "Overview of iRoC Technologies Report  
 "Radiation Results of the SER Test of Actel FPGA December 2005"

da: "Radiation Results of the SER Test of Actel, Xilinx and Altera FPGA instances"

nel nostro caso:  
 misure del 24/10 a 40nA a3p250  
 definisco:  
 cross section per chip:  $\text{Sigma\_C} = \text{No\_of\_errors} / (\text{fluence} * \text{C})$  with C= number of chips  
 reference fluence: fluence at ground level in New York City: fluence\_NYC = 14 n/cm^2/hour  
 FIT: Failure In Time: failures in 10^9 hours. FIT = Sigma\_C \* fluence\_NYC \* 10^9

misure del 24/10 a 40nA a3p250 Fifo errors in 1 hour(avg):  
 definisco:  
 cross section per FIFO Mb:  $\text{Sigma\_Mb} = \text{No\_of\_errors} * 10^6 / (\text{fluence} * \text{N\_FIFO\_bit})$   
 ref\_fluence 1,400E+01  
 FIT 5,460E+00

misure del 24/10 a 40nA a3p250 FlipFlop errors in 1 hour(avg):  
 definisco:  
 cross section per M flipflops:  $\text{Sigma\_Mff} = \text{No\_of\_errors} * 10^6 / (\text{fluence} * \text{N\_flipflops})$   
 ref\_fluence 1,400E+01  
 FIT 2,375E+02

our test at LNL	
FIT per million bit of embedded RAM	IROC test result per confronto
1,646E+03	1580
FIT per million flip flop	IROC test result per confronto
2,375E+02	889
ratios RAMbit errors/FF errors	1,77727784

**If our estimation of neutron flux at the CN facility in Legnaro are correct then for each 1/2 hour run at 40nA the FPGA (assuming 1cm^2 area) has been irradiated with ≈ order of 1\*10^10 neutrons.**

**So, according to current background simulations results (see Valentina Santoro's <http://www.fe.infn.it/~santoro/SuperB/Background/Neutrons/Touschek.html>), we got a few upset with neutron fluxes corresponding to many months of SuperB operation at the locations where the FPGA will sit.**



addressing the radiation issues:

measurement of TID and SEE on the optional front end op-amps and the EASIROC at the INFN Laboratori Nazionali di Legnaro (being planned)

USER GUIDE

*Omega*

## EASIROC

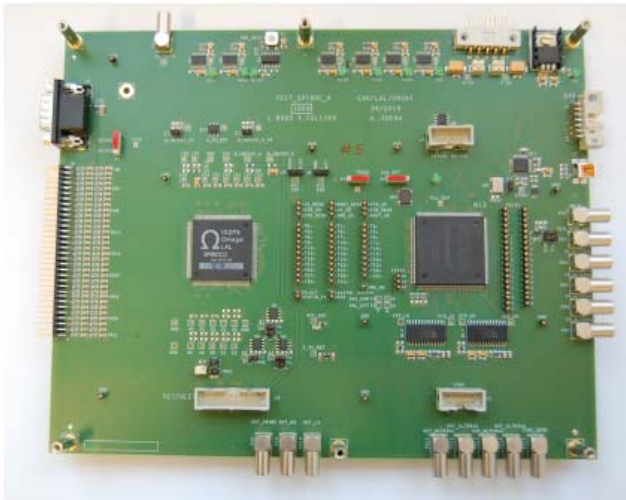
SOFTWARE & TEST BOARD USER GUIDE

Version: 11 April 2011

### Abstract

EASIROC (previously SPIROC), standing for *Extended Analogue Silicon pm Integrated Read Out Chip*, is a 32-channel fully-analogue front end ASIC dedicated to the gain trimming and read-out of SiPM detectors.

This guide explains how to install & use the test board for EASIROC and how to operate with the associated software.



A new test is foreseen at the neutron test facility in Legnaro toward the end of this month. The test is meant to evaluate the SEE rate in the DACs of the EASIROC chip and evaluate the TID dose effects on the analog signal processing units of the ASIC.

We also plan to look for TID effects on the op amps which could be used for the active “pick-up” option mentioned above.

We would stack the ACTEL FPGA board on top of the OMEGA board, in such a way to have correlated measurements failure rate measurements (the FPGA acting as a radiation monitor). In such a way the FPGA would act as some sort of neutron dose monitor, to be checked against the calibration data from the Legnaro facility.

  
*Orsay MicroElectronics Group Associated*