# HWP differential optical load and non-linearity

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- We calculate how much Popt variation is expected in each band due to HWP differential emissivity and transmission
- We simulate how the TES non-linearity up-modulates the  $2f_{_{HWP}}$  signal resulting in a I  $\rightarrow$   $4f_{_{HWP}}$  leakage
- We implement a simplified model to measure and remove the non-linearity

- Nominal power values → worse value in each band for a conservative estimation (inductive axis, except for 195 GHz and 402 GHz bands)
- Band-averaged estimation
- MHFT optical power variation wrt nominal values, due to HWP differential transmittance and emissivity







# Example: HWP differential emission @140 GHz compared with galactic emission

$$I_{\nu}^{ind} = BB(\nu, 20K) \cdot \epsilon_{140}^{ind}$$

$$I_{\nu}^{cap} = BB(\nu, 20K) \cdot \epsilon_{140}^{cap}$$

$$\int_{1.75}^{1.50} \frac{1.75}{1.25} \frac{1.00}{1.25}$$

$$\epsilon_{140}^{cap} = 0.007$$

$$\epsilon_{140}^{ind} = 0.019$$





- Differential emissivity causes a loss of power wrt the nominal power in each band
- In most bands %∆P<sub>HWP</sub> > %∆P<sub>GAL</sub>
- Signal from the sky is modulated by a  $2f_{HWP}$  signal with amplitude  $\Delta P_{HWP}$

## Input signal simulation

nside = 1024 ; smoothing fwhm=0.5 deg, 24 h obs

CMBpol @140 GHz + 2f<sub>HWP</sub> + 1/f noise; no fg

Anti-aliasing filter has been applied:

sample\_rate = 152 Hz duration = 3600\*24 s

q = 8 (downsampling factor to be applied later)



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# Modeling I(P) from Tijmen code (MNTES)

```
'Voltage bias': 8e-07,
'T_c': 0.18,
'transition_width': 0.002,
'R_normal': 1.0,
'T_bath': 0.1,
'z_series': (0.1+0j),
'n_index': 3.6,
'tau_intrinsic': 0.033,
'V_bias': 8e-07}
```



# Scheme of the implementation

Simulations: TOAST/litebird\_sim



(1) - Modeling I(P) from Tijmen code (MNTES)



# (2) - I(P) is calculated in the P range using (1) spline



#### **IMPORTANT NOTE FOR SIMULATIONS**

If we simulate at 19 Hz we see a lot of aliased signal!



Correct procedure (as in the on-board computer):

- simulation at 152 Hz (20MHz/2^17)
- anti-aliasing filter
- downsampling by a factor 8



#### **Fitting NL parameters**



- At this step we <u>did not include the dipole</u> in the simulation to simplify the analysis
- We simulate a 24 h observation
- We parametrize the HWP signal + NL through <u>4 parameters</u> as: x = Acos(2ωt+φ)
   f = x + bx<sup>2</sup> + d (quadratic approximation of the NL)
- We run a MCMC to find the best fit of the reconstructed power

$$\sigma_{A} = 8 \cdot 10^{-7} \text{ pW}$$

$$\sigma_{\phi} = 0.05 \text{ arcmin}$$

$$\sigma_{b} = 5 \cdot 10^{-4} \text{ pW}^{-1}$$

$$\sigma_{d} = 9 \cdot 10^{-7} \text{ pW}$$

# **Correction of the NL:** (3) - P is reconstructed using a quadratic model function



In fact, if we apply the same procedure with a third order model function we eventually obtain:



## Fitting NL parameters with the dipole



- We then <u>include the dipole</u> in the simulation
- We parametrize the HWP signal + dipole + NL through <u>5 parameters</u> as:  $x = Acos(2\omega t + \phi) + A_d dip(t)$

 $f = x + bx^2 + d$  (quadratic approximation of the NL)

$$\sigma_{A} = 8 \cdot 10^{-7} \text{ pW}$$
  

$$\sigma_{\phi} = 0.05 \text{ arcmin}$$
  

$$\sigma_{Ad} = 2 \cdot 10^{-3}$$
  

$$\sigma_{b} = 5 \cdot 10^{-4} \text{ pW}^{-1}$$
  

$$\sigma_{d} = 9 \cdot 10^{-7} \text{ pW}$$

# Summary

- We estimate the power variation due to HWP differential emissivity and transmission (up to 44%)
- We include the 2f<sub>HWP</sub> signal in the simulated input signal
- We apply the Tijmen model for TES non-linearity (MNTES) to this total signal
- We fit a quadratic approximation of the MNTES on the 2f<sub>HWP</sub> signal
  - $\circ$  to be assessed if the procedure is affected by other 4f<sub>HWP</sub> effects
  - $\circ$   $\sigma_{\phi}$  = 0.05 arcmin
- We correct the  $I \rightarrow 4f_{HWP}$  leakage due to detector non-linearity
  - $\circ$  4f<sub>HWP</sub> residual < white noise level
- <u>On going</u>: maps in presence of detector non-linearity